

# CHAPTER 4

## ELECTROMECHANICAL INDICATING INSTRUMENTS



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# Classification of Instruments

## 1. Indicating Instruments

- **indicate** measured value at an instant of time
- by a pointer and a scale or digital (numerical) display
- instrument e.g.: ammeter, voltmeter, thermometer
- quantity e.g.: current (A), voltage (V), temperature (K)

## 2. Recording Instruments

- **record** measured value over a period of time
- instrument e.g.: chart recorders, X-Y recorder, plotters
- recording e.g.: I vs t, V vs t, T vs t, I vs V

## 3. Integrating instruments

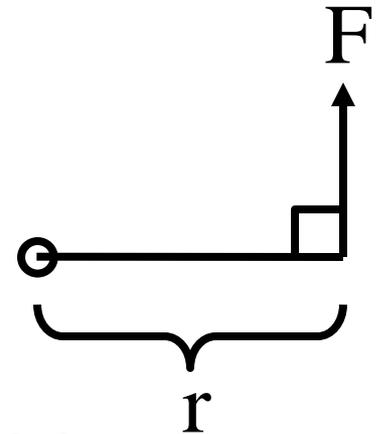
- **integrate** measured value over a period of time
- instrument e.g.: Watt-hour meter, ampere-hour meter
- quantity e.g.: energy (W-h), charge (A-h)

# Essentials of indicating instruments

All deflection instruments consist of a *pointer* attached to a *moving system* that moves the pointer over a *calibrated scale*.

The **moving system** is subjected by 3 torques:

- A deflecting torque
- A controlling torque
- A damping torque



**Torque = force × distance from center**

$$\mathbf{T = F \times r \text{ (N-m)}}$$

## Deflecting / Operating torque ( $T_d$ )

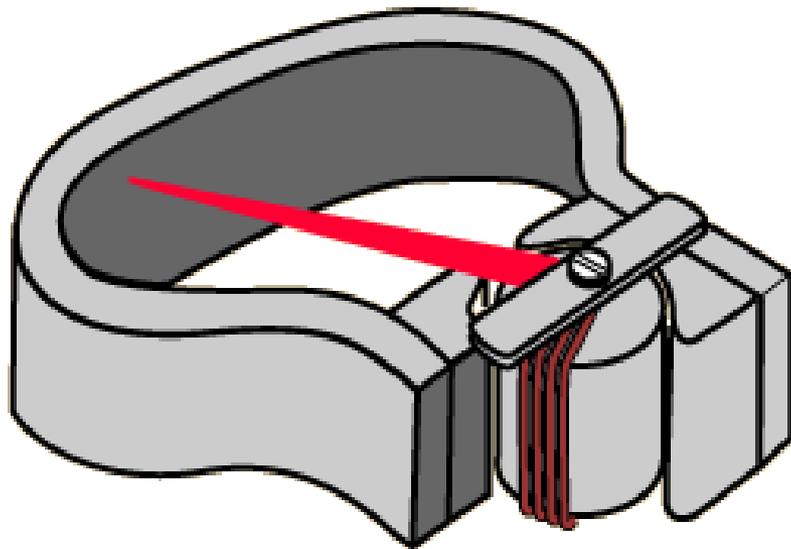
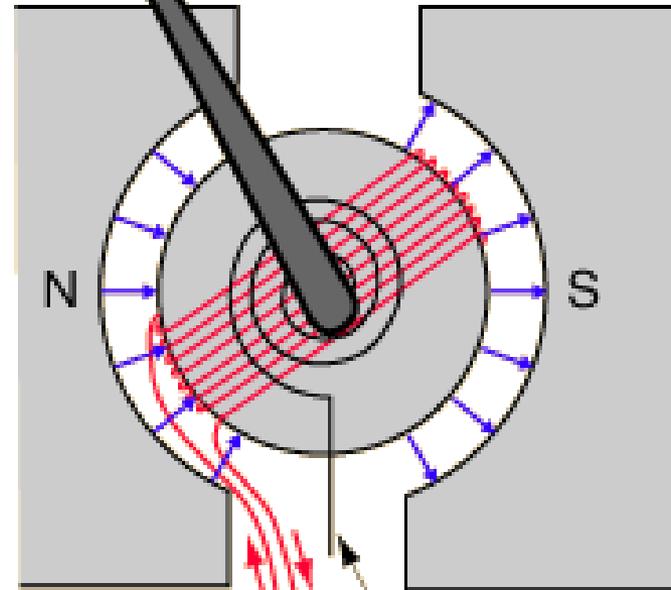
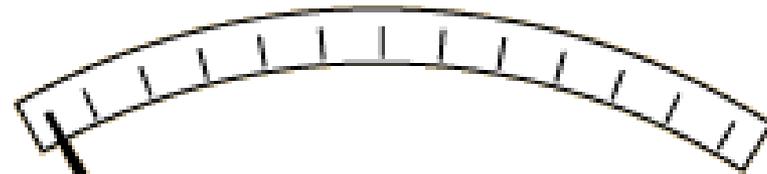
$T_d$  is produced using one of the below effects:

- electromagnetic
- electrodynamic
- inductive
- thermal
- electrostatic

1. All these effects can be related to electric current.
2.  $T_d$  causes the moving system to move from its zero position.
3. Some deflection instruments are totally not electrically related

e.g.: Bourdon tube *pressure meter*.

# GALVANOMETER – ELECTRIC CURRENT DETECTOR



Current  $I$

Restoring spring

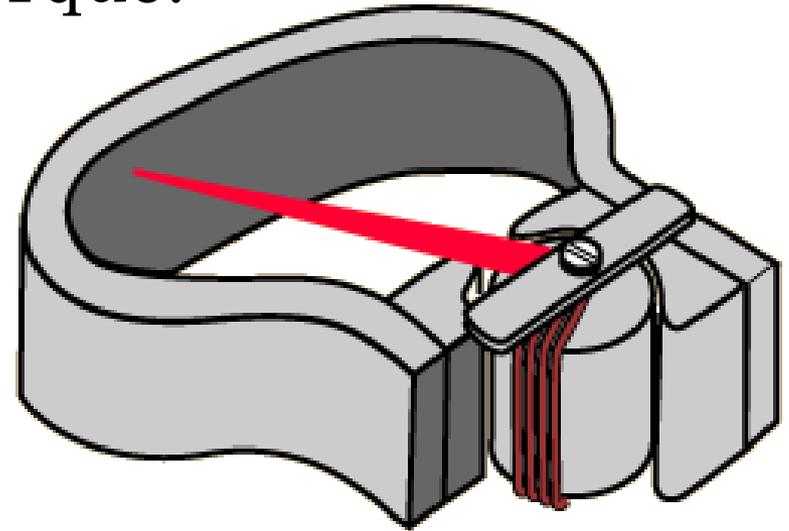
**PMMC**

# PMMC: PERMANENT MAGNET MOVING COIL (D'ARSONVAL MOVEMENT)

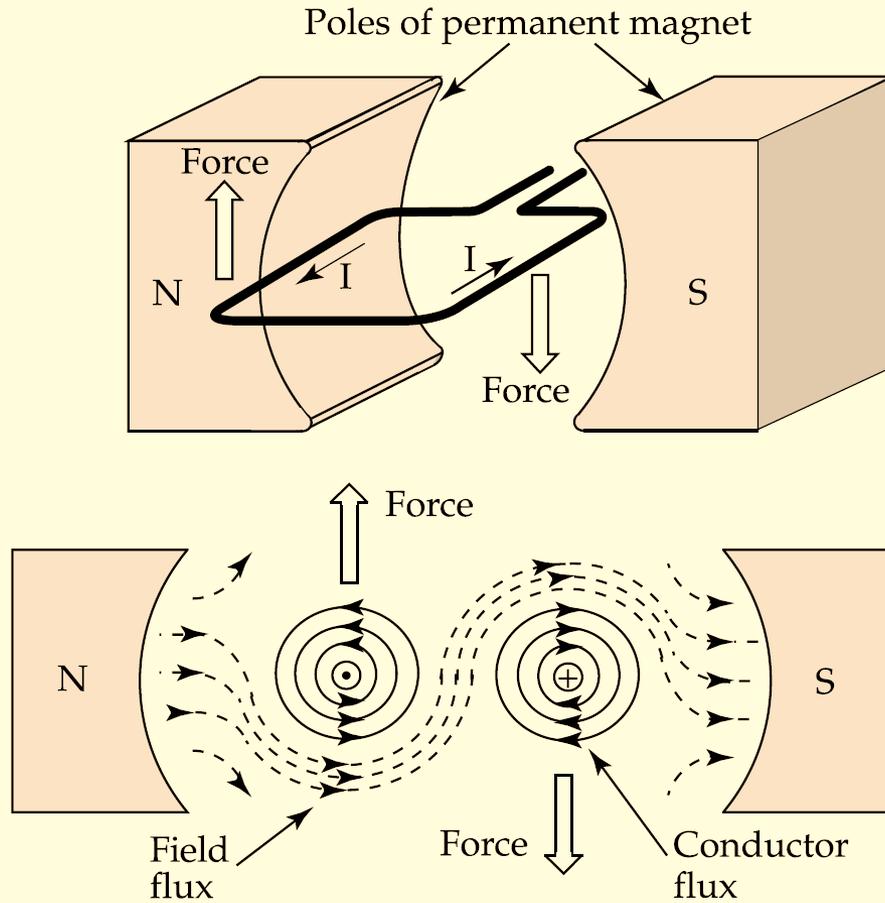
- The equation for the developed torque derived from the basic law for electromagnetic torque:

- $$T = B \times A \times I \times N$$

- **T**= Torque (N.m)
- **B**= Air-gap Flux density
- **A**= Effective coil area
- **I**= Current in the moving coil
- **N**= turns on the coil

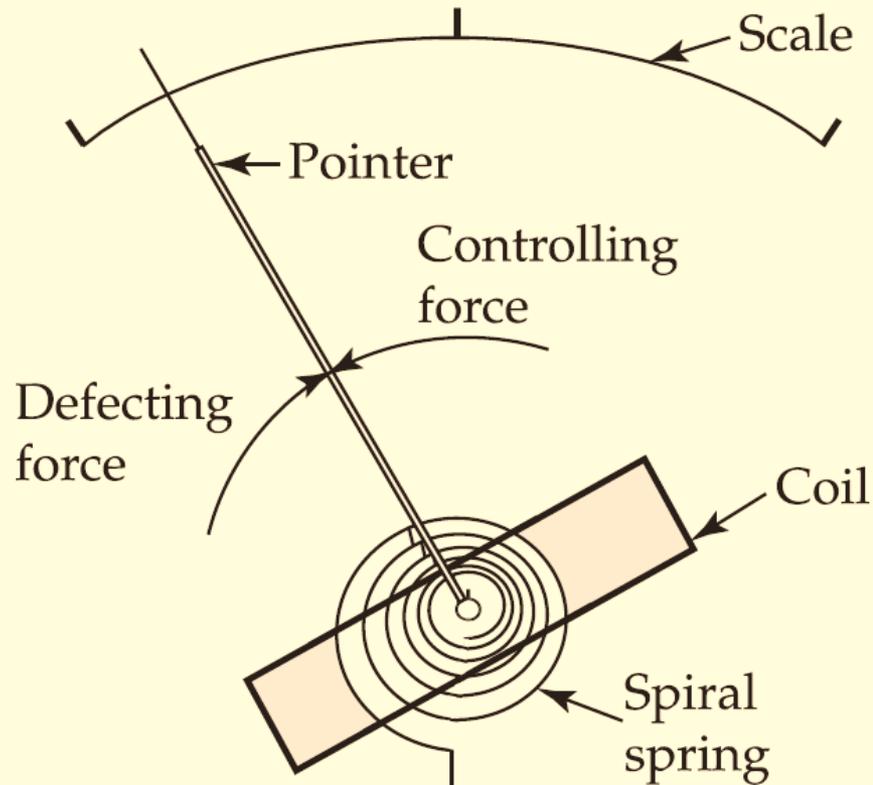


# Deflection Instrument Fundamentals



**Figure 3-1** The deflecting force in a PMMC instrument is produced by the current in the moving coil. This sets up a magnet flux which interacts with the flux from the poles of the permanent magnet.

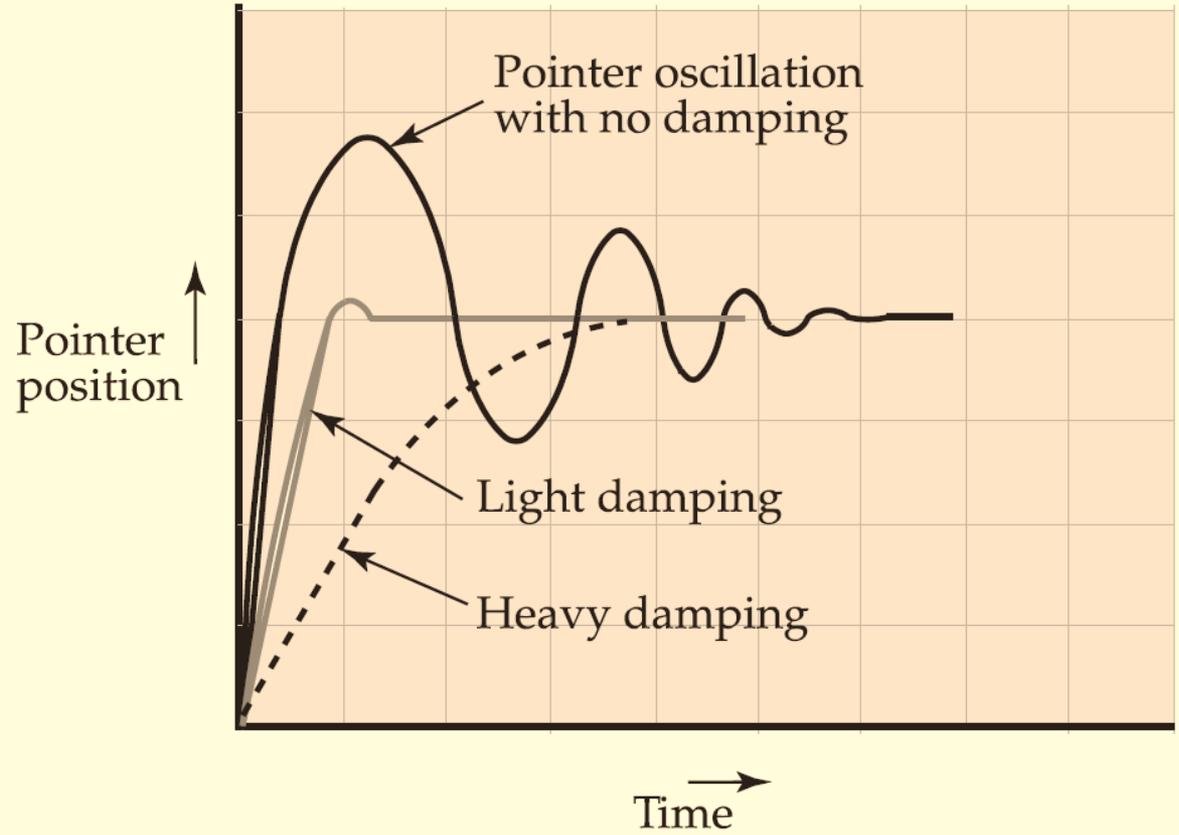
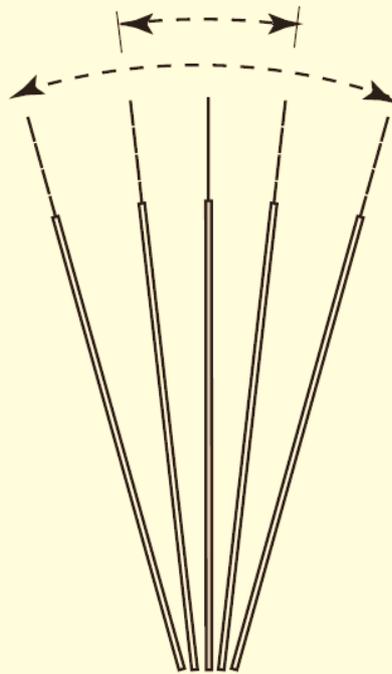
# Deflection Instrument Fundamentals



**Figure 3-2** The controlling force in a PMMC instrument is provided by spiral springs. The two forces are equal when the pointer is stationary.

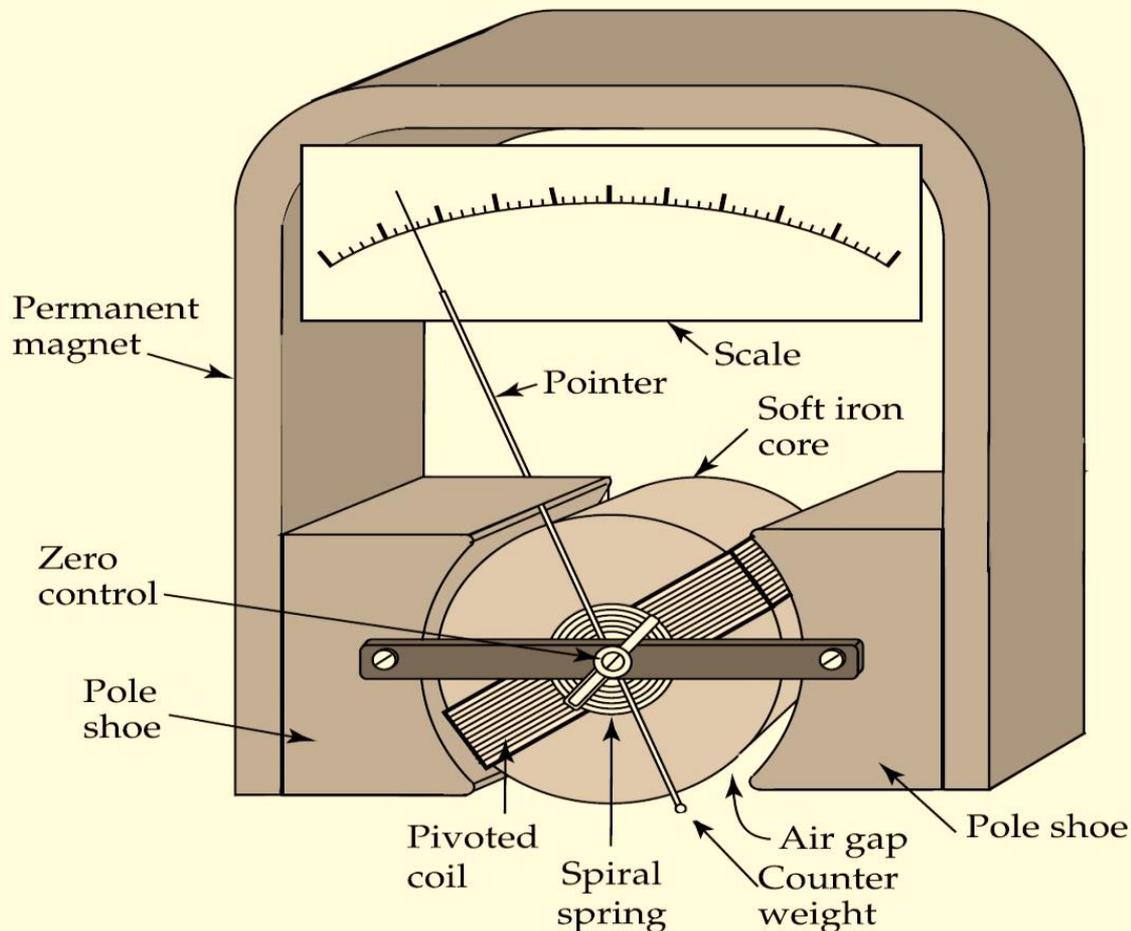
# Suspension

Pointer oscillation



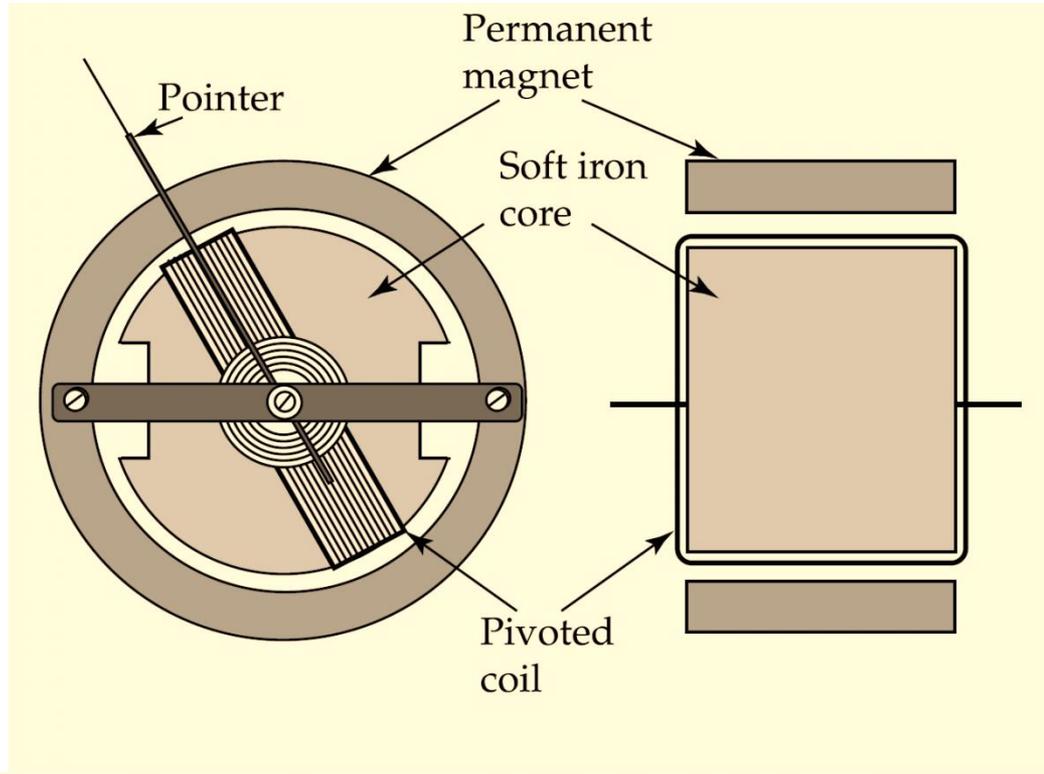
(a) Absence of damping force causes the pointer to oscillate

# Permanent Magnet Moving Coil Instrument



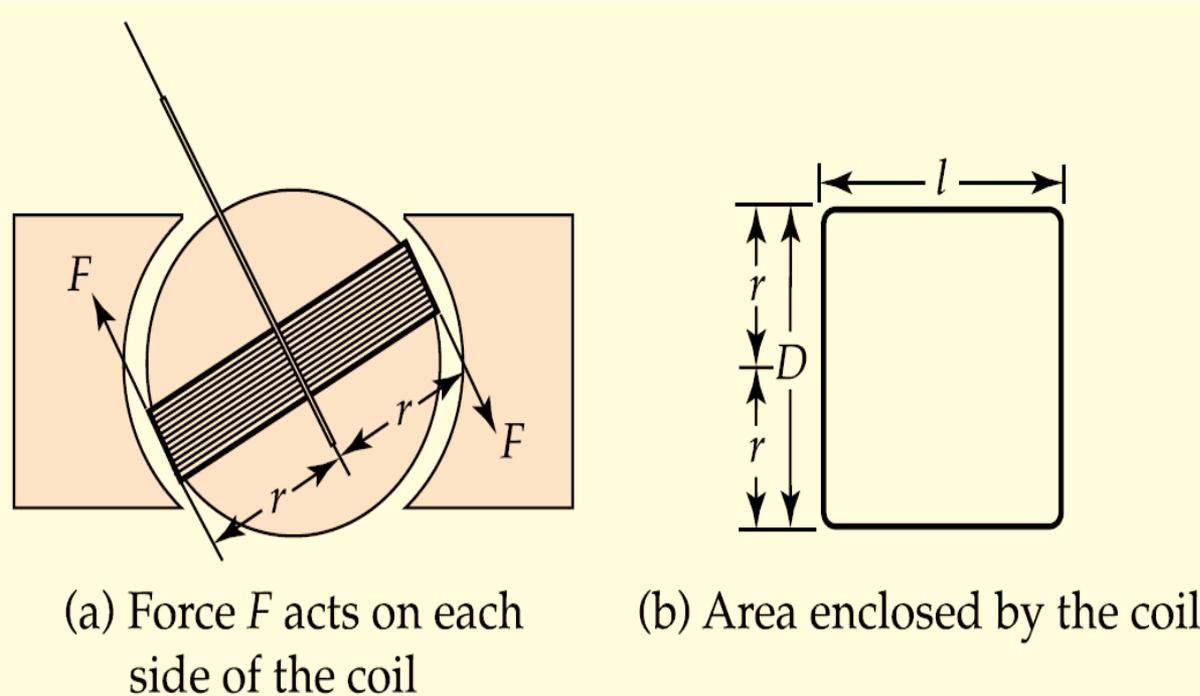
**Figure 3-6** A typical PMMC instrument is constructed of a horseshoe magnet, soft-iron pole shoes, a soft-iron core, and a suspended coil that moves in the air gap between the soft-iron core and the pole shoes.

# Construction of Permanent Magnet Moving Coil Instrument



**Figure 3-7** In a core-magnet PMMC instrument, the permanent magnet is located inside the moving coil, and the coil and magnet are positioned inside a soft-iron cylinder.

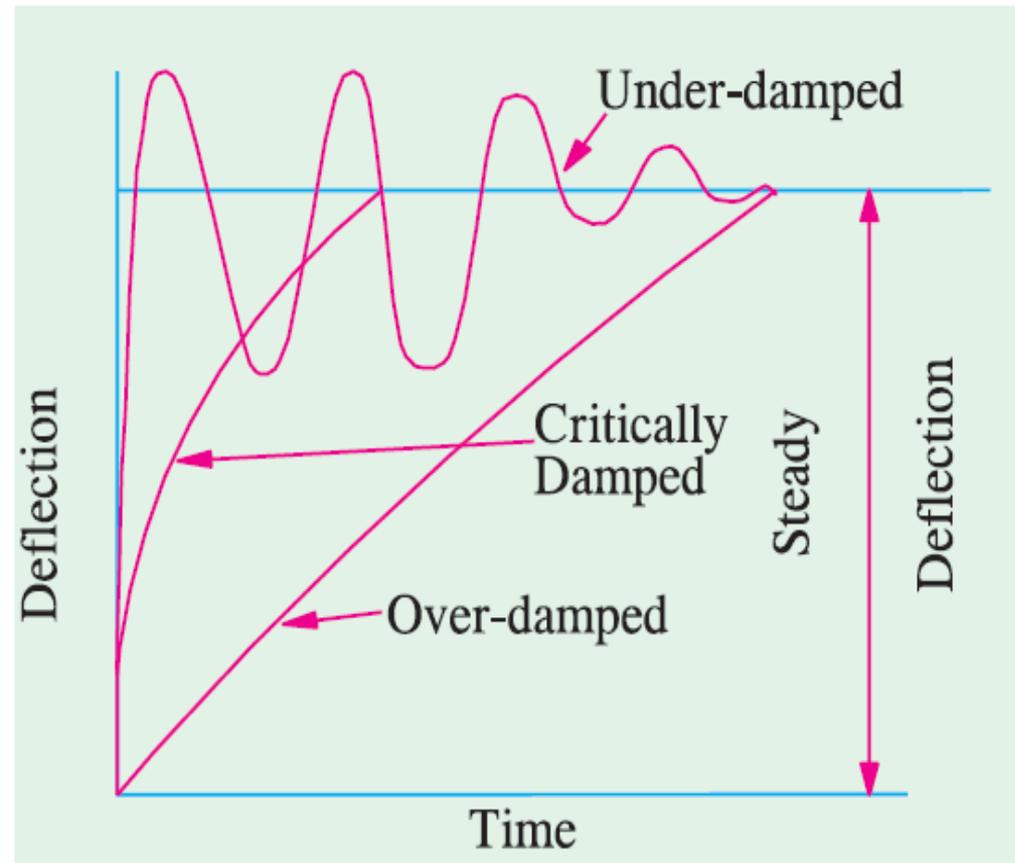
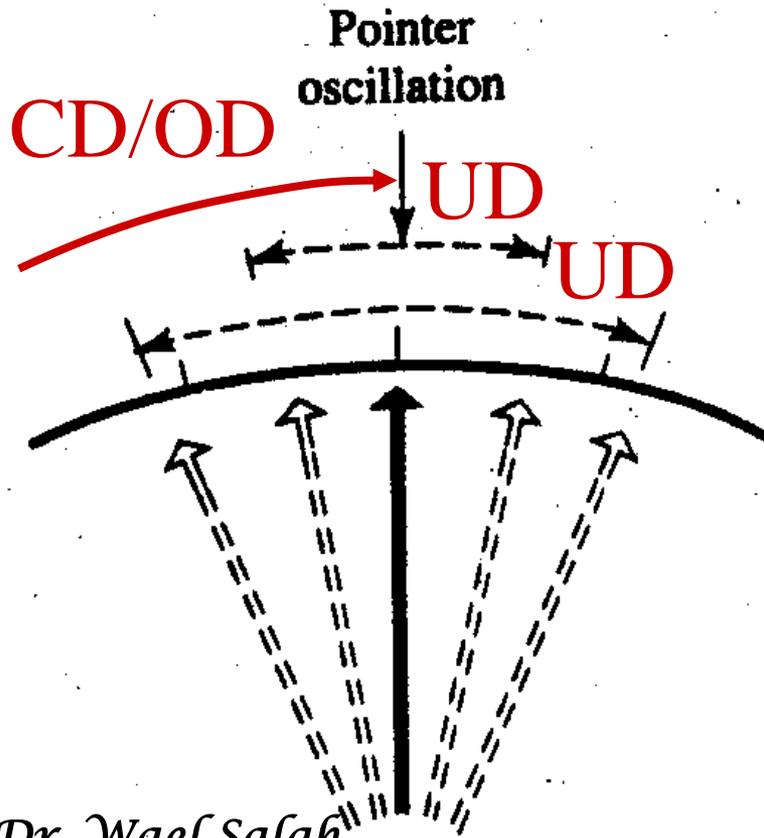
# Torque Equation & Scale



**Figure 3-8** The deflecting torque on the coil of a PMMC instrument is directly proportional to the magnetic flux density, the coil dimensions, and the coil current. This gives the instrument a linear scale.

# Damping Torque

- acts on the moving system only when it is moving and always **opposes its movement**
- Efficient damping: quickly reach final position without overshooting.

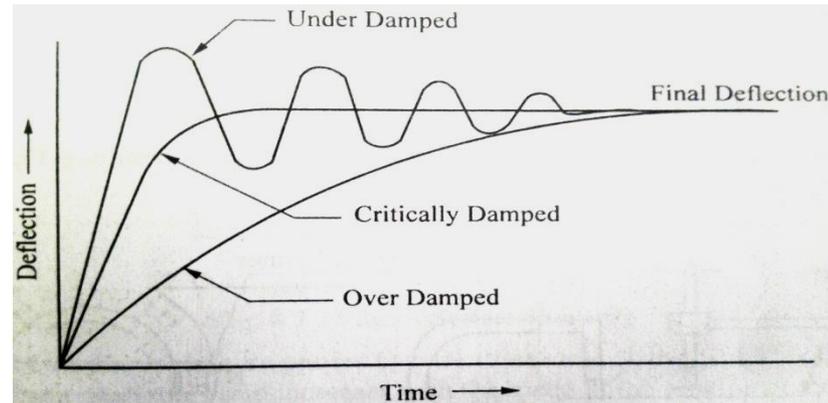


## Over Damping:

The coil returns **slowly** to its rest position without overshoot or oscillation.

## Under Damping:

The coil movement is subjected to **sinusoidal** oscillation.



## Critical Damping:

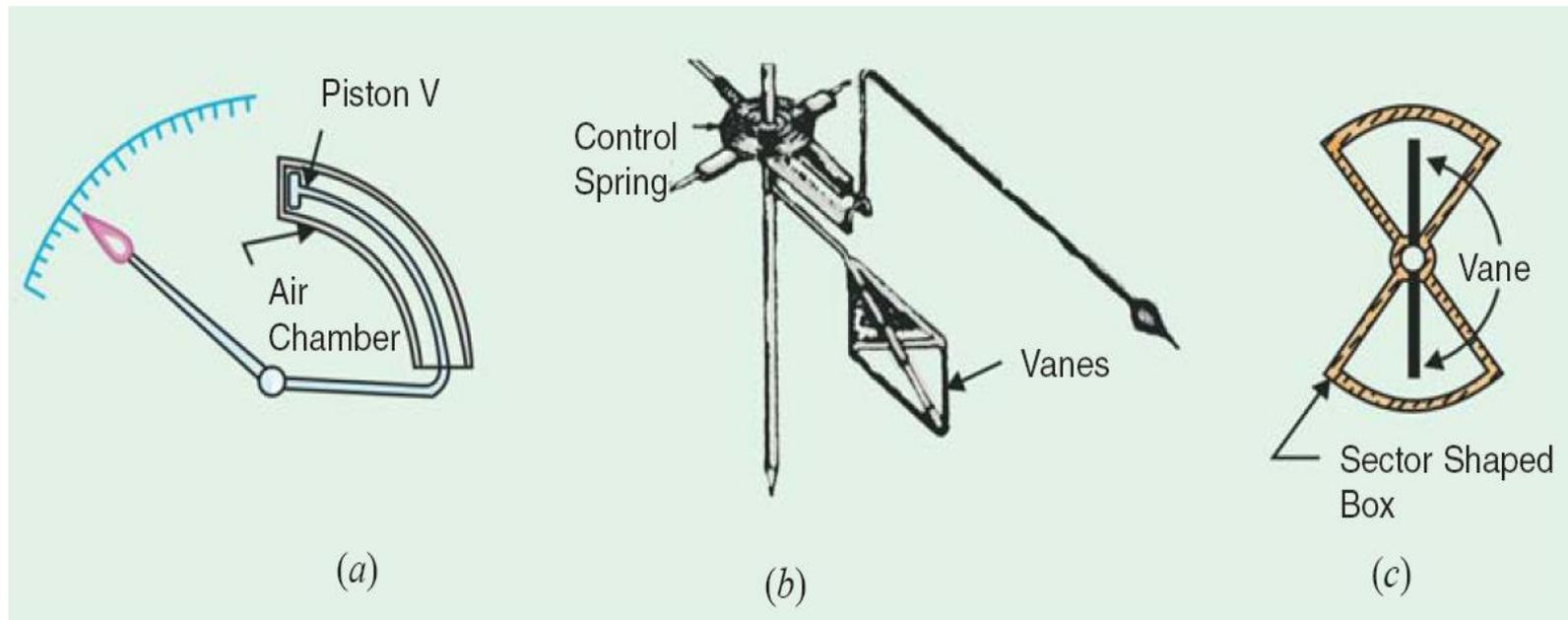
The coil returns **immediately** to its steady-state position without oscillation.

# *Method of Damping*

## 1- Mechanical

- **Air** friction (used if eddy current is not suitable)
- **fluid** friction (not often used)

## Air friction damping: Air-chamber and Vanes



# *Method of Damping*

## **2- Electromagnetic**

**Eddy** current (very efficient)

## **3- Electrical**

**CDRE**: Critical Damping Resistance External  
The CDRE connected in parallel with the coil.

## Controlling / restoring / balancing torque ( $T_C$ )

$T_C$  opposes  $T_d$

$T_C$  increases with deflection angle ( $\theta$ )

*When  $T_C = T_d$ , the moving system will be at rest.*

*When  $T_d$  is removed*, the moving system will be returned (restored) back its zero position by  $T_C$ .

*If  $T_C$  is not introduced to the moving system*, the moving system will move *continuously* over its maximum deflection position, as long as  $T_d > 0$ .

Two methods for  $T_C$ :

- A **spring** - spring control
- A **weight** - gravity control

# Methods for Controlling Torque ( $T_C$ ) .....

## 1) Spring control

A spirally wound hair-spring is used.

*When the spring is twisted from its equilibrium position, a restoring torque ( $T_C$ ) is produced.*

### Example:

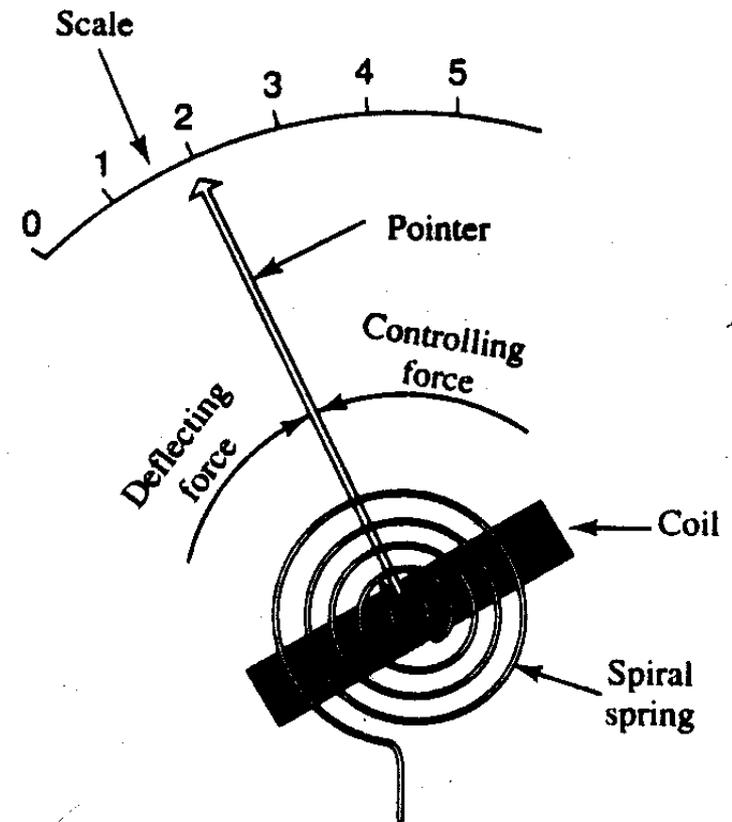
*Permanent-magnet moving coil (PMMC) instruments:*

$$T_d \propto I$$

However,  $T_C \propto \theta$

At  $T_C = T_d$ ,  $\theta \propto I$

or  $I \propto \theta$  (the scale is linear)



Methods for  $T_C$  .....

## **Spiral spring characteristics:**

The number of turns is fairly large so that no deformation on the spring occurs. Then

$$T_C = K\theta \quad \text{for } 0 \leq \theta \leq \theta_{\max}$$

$K$  = spring constant (N-m/degree)

$\theta$  = deflection angle from  $T_C = 0$  position.

## **Materials to make the spring must be:**

- non-magnetic
- not subject to much deterioration with time
- low temperature coefficient of spring constant

*Hence, phosphor-bronze material is used.*

## 2) Gravity control (**seldom** used nowadays)

Based on adding some **weights** to control the movement of the indicator.

# Temperature Compensation

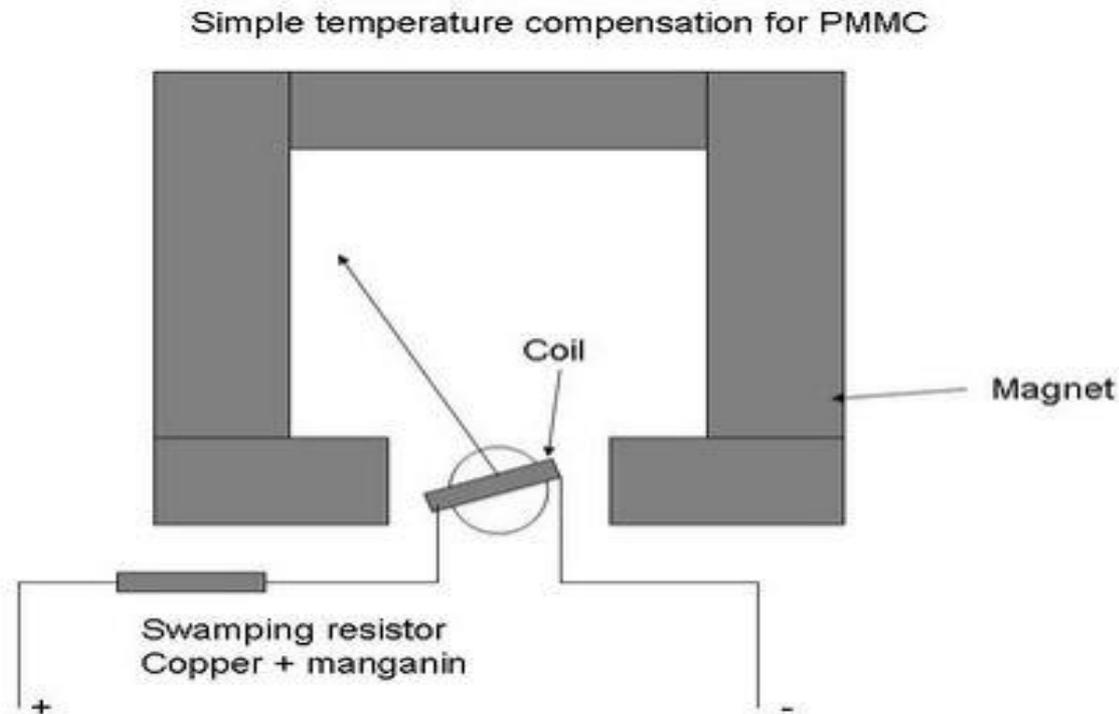
- **Magnetic Field strength** decrease with temp
- **Spring Tension** decrease with temp
- **Coil Resistance** increase with temp.

These factors causes the pointer to read low for a given current with respect to magnetic field strength.

**The temperature may be compensated** by appropriate use of series and shunt resistors with the moving coil.

## Temperature Compensation .....

The simple temperature compensation circuit for PMMC uses a resistance in series with a movable coil, as shown in the figure. The resistor is called swamping resistor. It is made up of manganin having practically zero temperature coefficients, combined with copper in the ratio of 20/1 or 30/1.



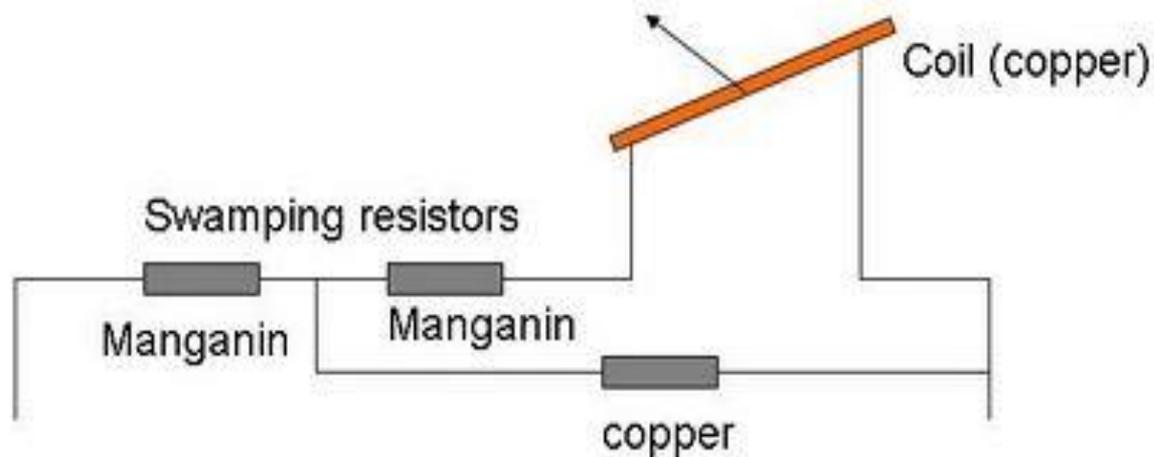
The resultant resistance of the coil and the swamping resistor increases slightly as temperature increases, just enough to compensate the change in springs and magnet due to temperature. Thus the effect of temperature is compensated.



## Temperature Compensation .....

More complicated but complete cancellation of temperature effects can be obtained by using the swamping resistors in series and parallel combination as shown in figure.

### Improved temperature Compensation



In this circuit, by correct proportioning of copper and manganin parts, complete cancellation of the temperature effects can be achieved.

# **INDICATING INSTRUMENTS FOR CURRENT AND VOLTAGE MEASUREMENTS**

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*First Exam on*  
**30/03/2015**

# Moving Coil Instruments as DC or AC ammeters & voltmeters

## 1. *Permanent magnet moving coil (PMMC)*

- a moving coil & a pair of permanent magnets  
(basically for **DC** measurements)

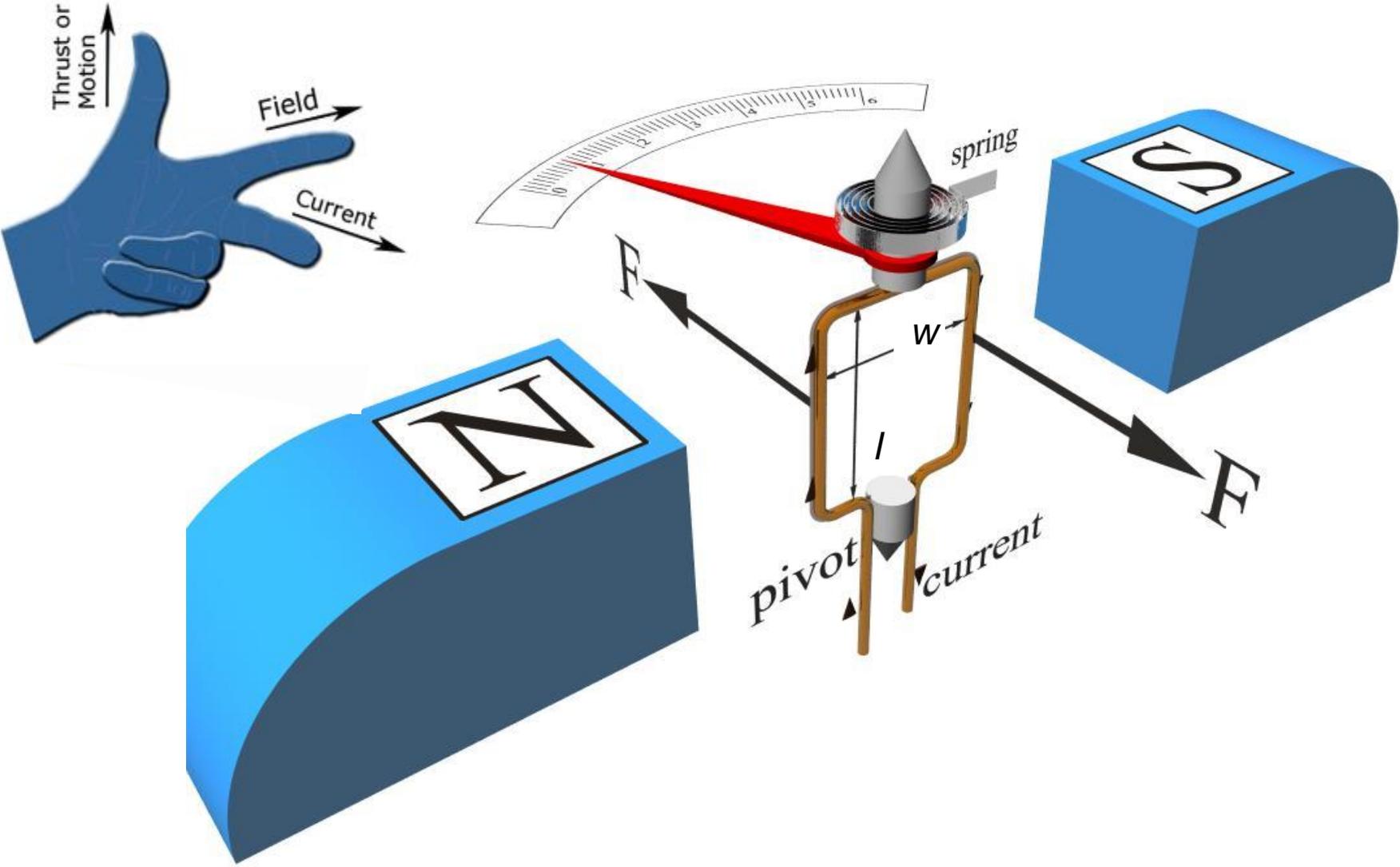
## 2. *Dynamometer or electrodynamic*

- a moving coil and a pair of fixed coils  
(for **DC** and **AC** measurements)

**PMMC** - low  $N$  (hence low  $L_{\text{coil}}$ ),  $R_{\text{coil}}$  and  $V_{\text{drop}}$

**Dynamometer** - large  $N$ , moderate  $R_{\text{coil}}$  and  
moderate  $L_{\text{coil}}$

# PMMC



Reference: [http://physbin.com/portfolio/imgs2/index\\_imgs\\_topic-051mag-g-elecdyn.htm](http://physbin.com/portfolio/imgs2/index_imgs_topic-051mag-g-elecdyn.htm)

# Current range extension

DC Ammeters: by a **shunt** (low R)

*Multiplying power* (**Meter current very low**)

$$I_m = I_{FSD}$$

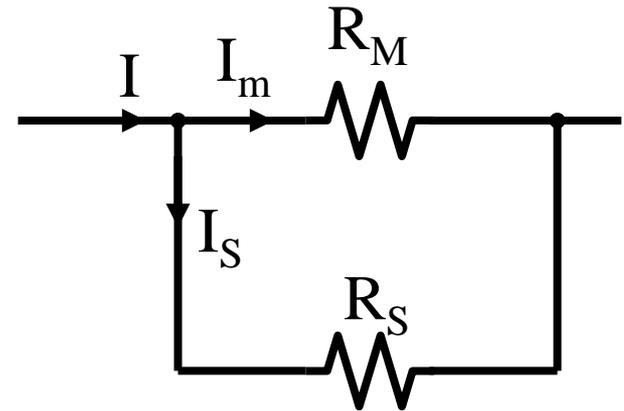
$$V_{shunt} = V_{movement}$$

$$I_s R_s = I_m V_m$$

$$R_s = \frac{I_m R_m}{I_s}$$

$$I_s = I - I_m$$

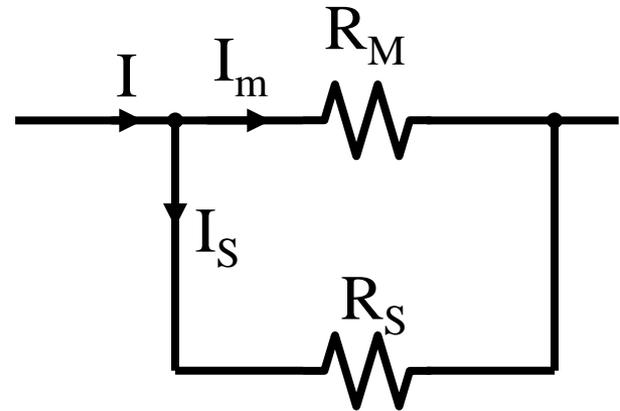
$$R_s = \frac{I_m R_m}{I - I_m}$$



## *Example:*

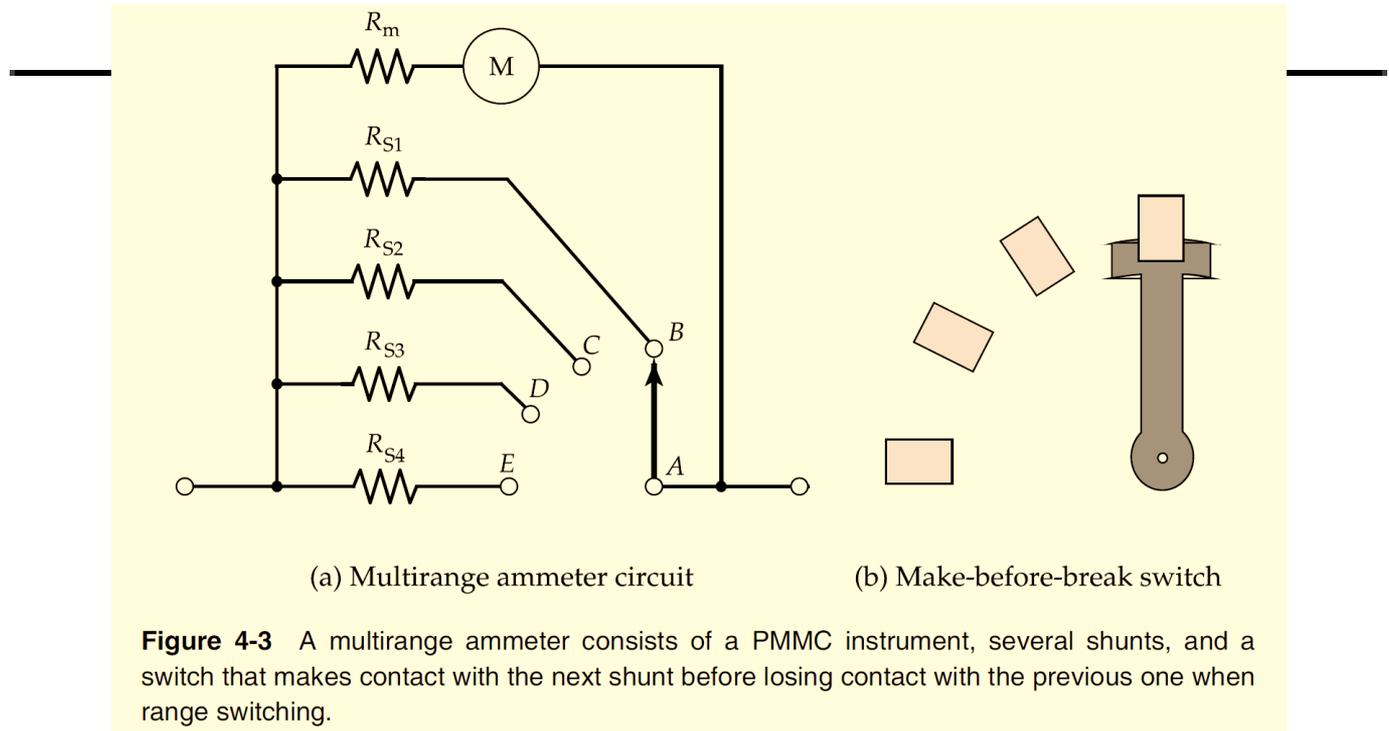
**A 1 mA meter movement with internal resistance of 100 ohm, is to be converted into a 0-100 mA ammeter. Calculate the value of shunt resistor required.**

$$R_s = \frac{I_m R_m}{I - I_m} = \frac{1 \times 10^{-3} \times 100}{100 \text{ mA} - 1 \text{ mA}} = 1.01 \Omega$$



# Current range extension

- \* The current range of a dc ammeter can be further extended by range switch to form multiage ammeter.
- $R_1, \dots, R_4$  gives different current ranges.

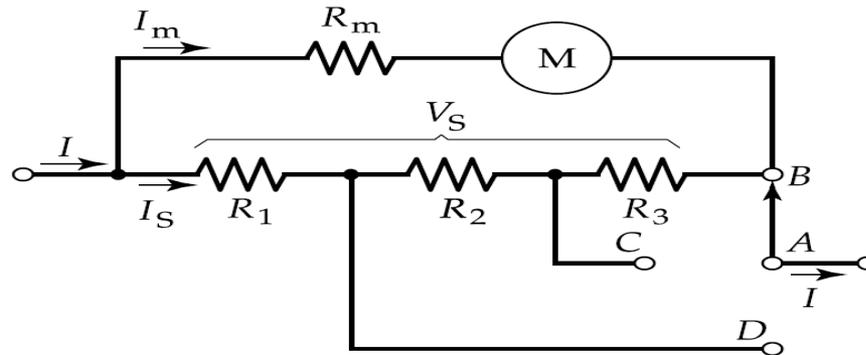


For this type: During changing the shunt by selector switch the meter is connected **directly without shunt**  
➔ This may cause the damage of the meter.

# Ayrton (Universal) Shunt

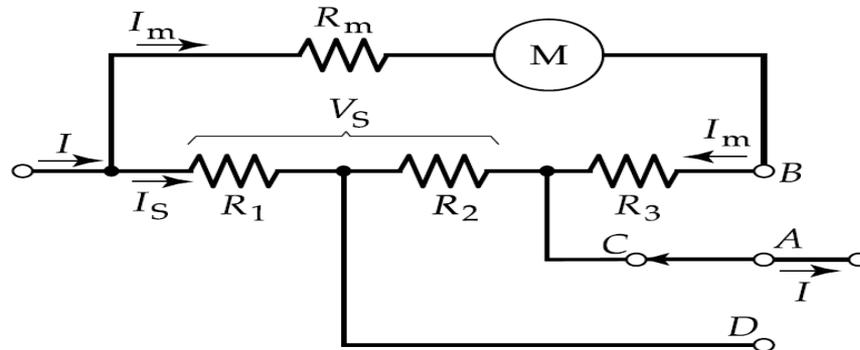
→ This connection will eliminate the possibility of having the meter circuit without shunt.

→ For higher current range the shunt resistor should be lower



(a) Ayrton shunt and meter

$$(R_1 + R_2 + R_3) \parallel R_m$$



(b) Switch at terminal C

$$(R_1 + R_2) \parallel (R_m + R_3)$$

## *Example:*

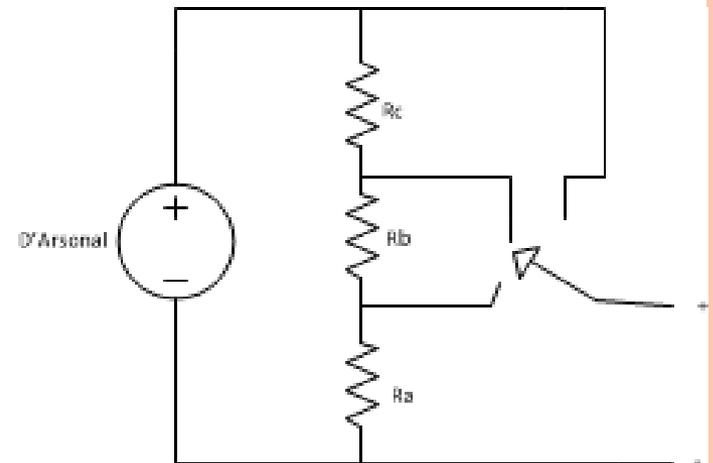
**Design an Ayrton shunt Ammeter using D'Arsonval movement to measure current with ranges of 1A, 5A and 10A. Given that the meter internal resistance  $R_m=50\ \Omega$  and the Full Scale Deflection current is 1 mA.**

$$I_s = I - I_m = 1A - 1mA = 999\ mA$$

$$R_a + R_b + R_c = \frac{1 \times (50\ \Omega)}{999} = 0.050005\ \Omega$$

$$R_a + R_b = \frac{1 \times (R_c + 50\ \Omega)}{4.999}$$

$$R_a = \frac{1 \times (R_b + R_c + 50\ \Omega)}{9.999}$$

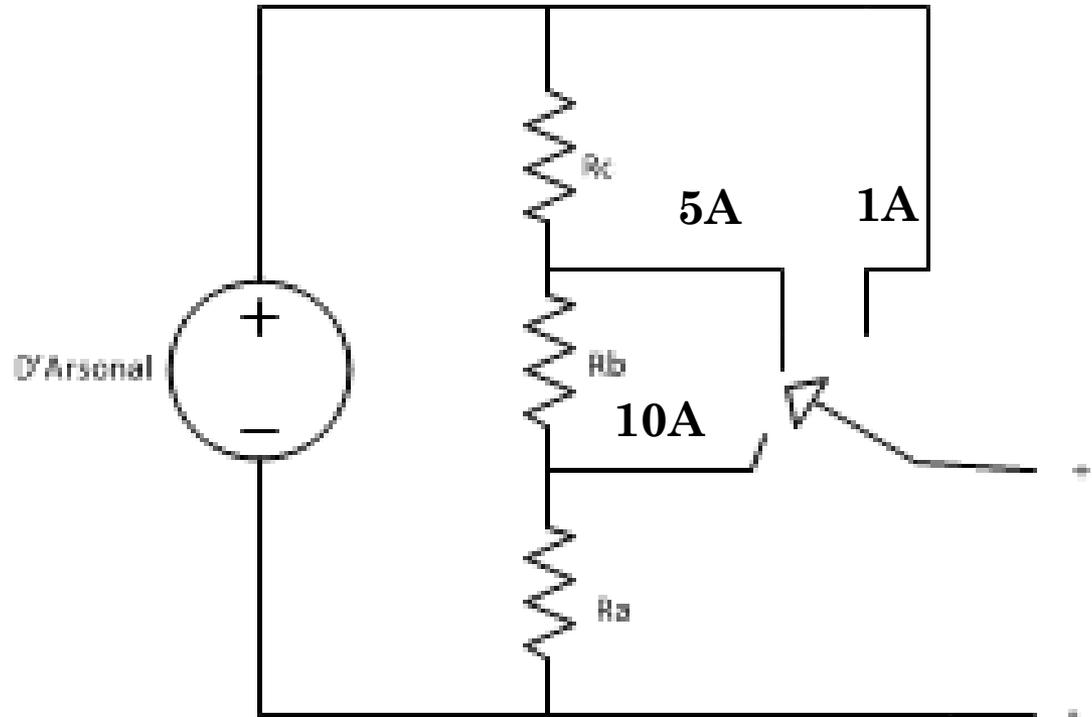


**Solving Equations 1, 2 and 3 giving:**

$$R_a = 0.0050005\Omega$$

$$R_b = 0.0050005\Omega$$

$$R_c = 0.04004\Omega$$



**For lager current value we need smaller resistor**

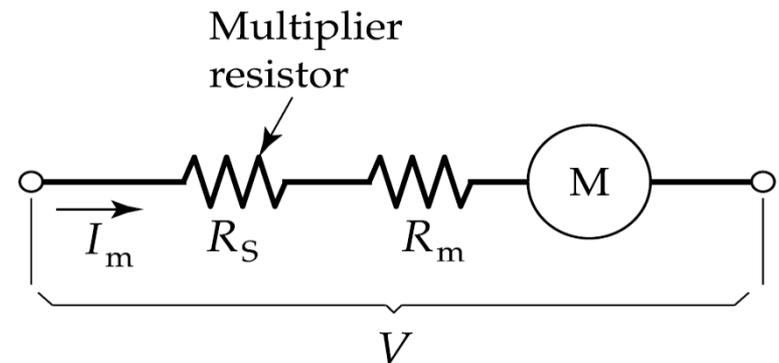
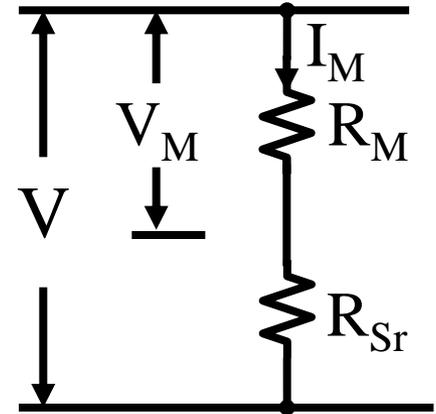
# Voltage range extension

DC Voltmeters: by a **multiplier** (high R)

*Voltage magnification*

$$V = I_m (R_s + R_m)$$

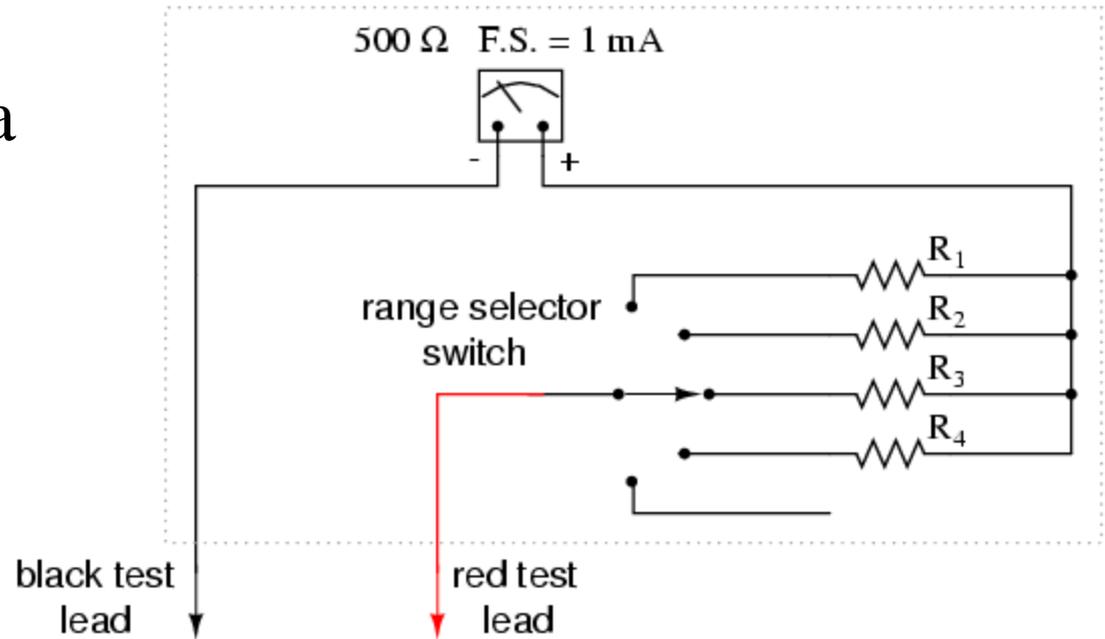
$$R_s = \frac{V - I_m R_m}{I_m} = \frac{V}{I_m} - R_m$$



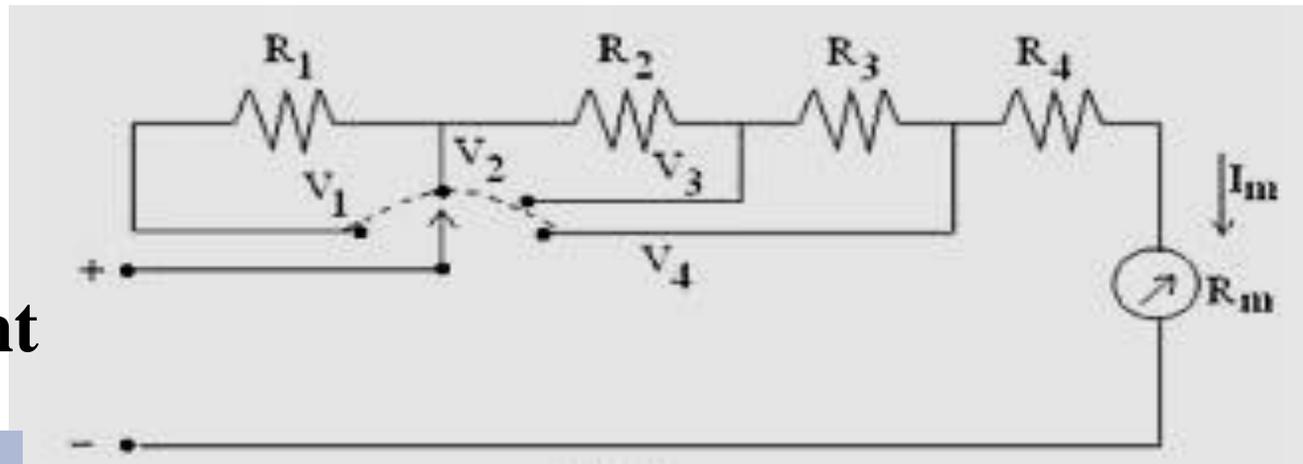
# Multi range Voltmeter

DC Voltmeters: by a **multiplier** (high R)

*A multi-range voltmeter*

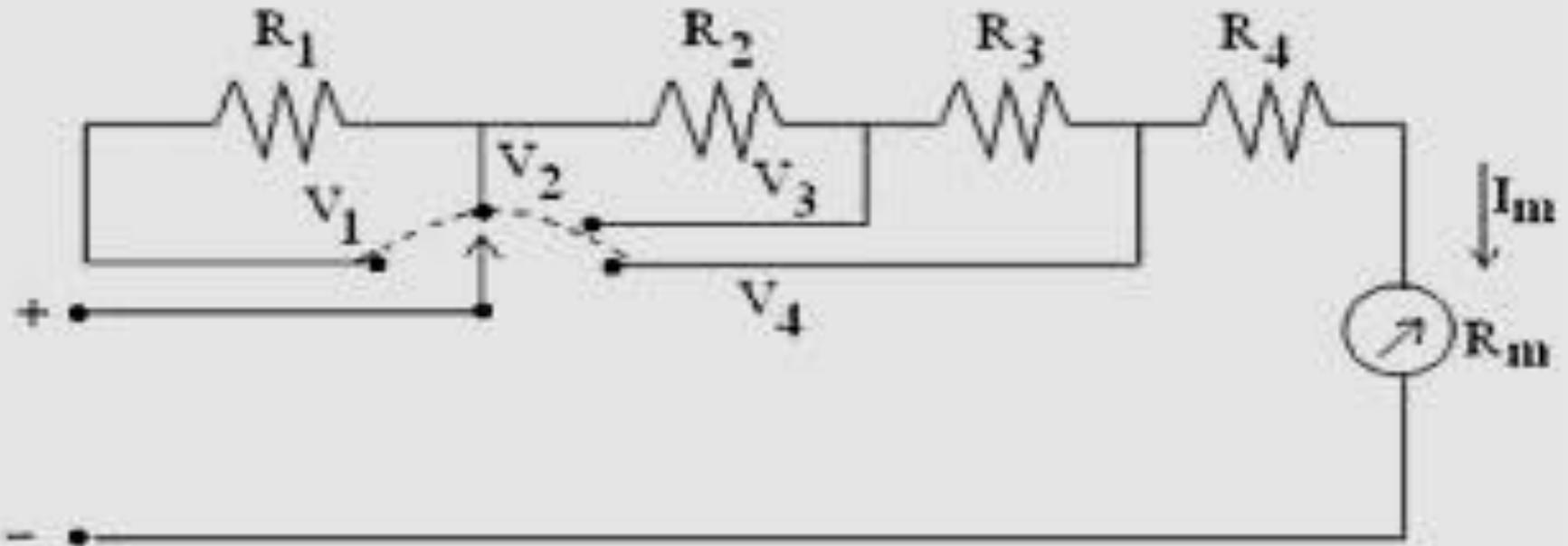


More practical arrangement



## More practical arrangement

- Resistors are connected in series
- Advantage: All multiplier resistor **except 1<sup>st</sup> ( $R_4$ )** one have **standarded** resistor values.



**Example:**

A basic D'Arsonal movement with an internal resistance of  $100\Omega$ , and full scale current of  $1\text{mA}$ , to be converted into a multirange dc voltmeter with series connected resistor to measure voltages ranges  $10\text{V}$ ,  $50\text{V}$ ,  $250\text{V}$  and  $500\text{V}$ .

10 V (posisi V4)

$$R_T = \frac{10\text{V}}{1\text{mA}} = 10\text{ k}\Omega$$

$$R_4 = R_T - R_m = 10\text{ k}\Omega - 100\Omega = 9,900\Omega$$

50 V ( posisi V3 )

$$R_T = \frac{50\text{V}}{1\text{ mA}} = 50\text{ k}\Omega$$

$$R_3 = R_T - (R_4 + R_m) = 50\text{ k}\Omega - 10\text{k}\Omega = 40\text{k}\Omega$$

250 V (posisi V2)

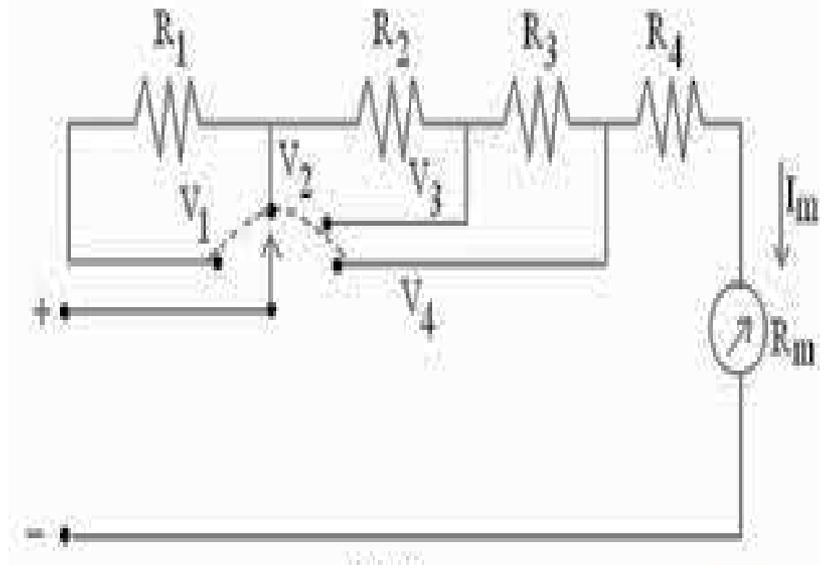
$$R_T = \frac{250\text{V}}{1\text{ mA}} = 250\text{ k}\Omega$$

$$R_2 = R_T - (R_3 + R_4 + R_m) = 250\text{k}\Omega - 50\text{k}\Omega = 200\text{k}\Omega$$

500 V ( posisi V1)

$$R_T = \frac{500\text{V}}{1\text{ mA}} = 500\text{k}\Omega$$

$$R_1 = R_T - (R_2 + R_3 + R_4 + R_m) = 500\text{ k}\Omega - 250\text{k}\Omega = 250\text{k}\Omega$$



# Voltmeter Sensitivity

Voltmeter Sensitivity: ratio of the **total resistance / range of voltage**

Example  $10\text{K}\Omega / 10\text{V} = \mathbf{1000 \Omega / V}$  (for last example)

$$S = \frac{1 \Omega}{I_{fsd} V}$$

**S = Sensitivity of the voltmeter**

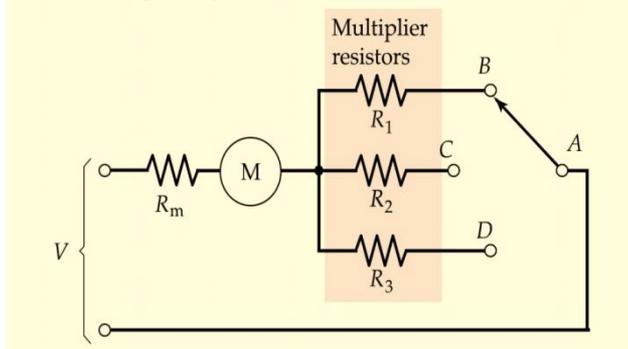
**V = Voltage range**

**R<sub>m</sub> = Movement internal resistance + R<sub>4</sub>**

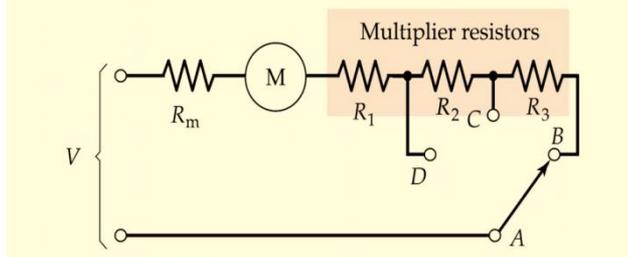
**R<sub>s</sub> = Resistance of the multiplier.**

$$R_T = S \times V$$

$$R_S = (S \times V) - R_m$$



(a) Multirange voltmeter circuit using switched multiplier resistors



(b) Multirange voltmeter circuit using series connected multiplier resistors

**Example:** with reference to the previous example

$$S = \frac{1}{I_{fsd}} = \frac{1}{0.001A} = 1,000 \frac{\Omega}{V}$$

**R4**

$$R_S = \left(\frac{1000\Omega}{V}\right) \times 10V - 100\Omega = 9,900\Omega$$

**R3**

$$R_S = \left(\frac{1000\Omega}{V}\right) \times 50V - 10,000\Omega = 40k\Omega$$

**R2**

$$R_S = \left(\frac{1000\Omega}{V}\right) \times 250V - 50k\Omega = 200k\Omega$$

**R1**

$$R_S = \left(\frac{1000\Omega}{V}\right) \times 500V - 250k\Omega = 250k\Omega$$

# Voltmeter Loading Effect

Low Sensitivity meter may give correct reading when measuring voltage in low resistance circuits.

But it produce unreliable reading in high resistance circuits.

The voltmeter acts as a **SHUNT** for that portion of the circuit

→ thus reduces the equivalent resistance

→ gives a lower indication of the voltage

*This called loading effect*

Thus voltmeter with low sensitivity gives higher error

# Voltmeter Loading Effect

The voltage across a **50kΩ** resistor in a circuit. Two voltmeters are available for measurement. Voltmeter 1 with sensitivity **1,000Ω/V** and voltmeter 2 with sensitivity **20,000Ω/V**. Both meters are used on their **50V range**.

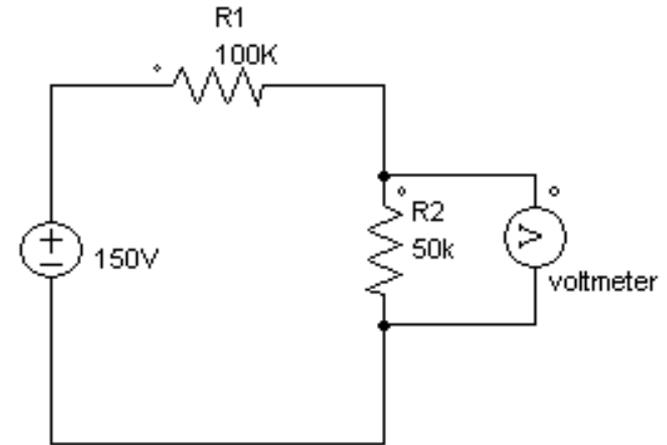
- Calculate the reading for each meter.
- Calculate the %Error in each reading.

**Solution:**

$$V_{R2} = \frac{50k\Omega}{150k\Omega} \times 150V = 50V_{True\_Value}$$

$$V_{R2} = \frac{25k\Omega}{125k\Omega} \times 150V = 30V_{Meter1}$$

$$V_{R2} = \frac{46.6k\Omega}{146.6k\Omega} \times 150V = 48.36V_{Meter2}$$



$$R_{meter} = V_{range} \times S$$

$$R_{meter} \parallel R_2$$

$$\text{- \% Error 1} = \frac{50-30}{50} \times 100\% = 40\%$$

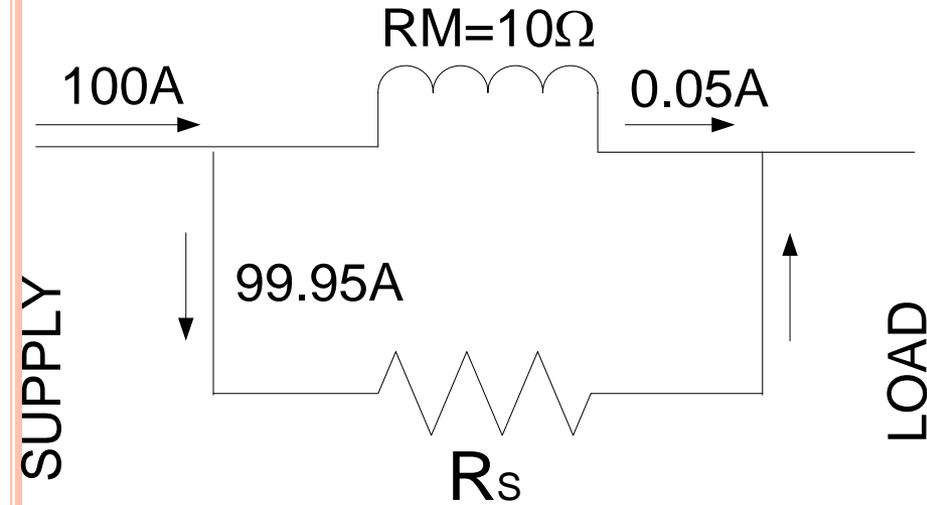
$$\text{- \% Error 2} = \frac{50-48.36}{50} \times 100\% = 3.28\%$$

## *EXAMPLE:*

A moving-coil instrument has a resistance of  $10\Omega$  and gives full-scale deflection when carrying a current of 50-mA.

Show how it can be adopted to measure voltages up to 750V and currents up to 100A.

## SOLUTION:



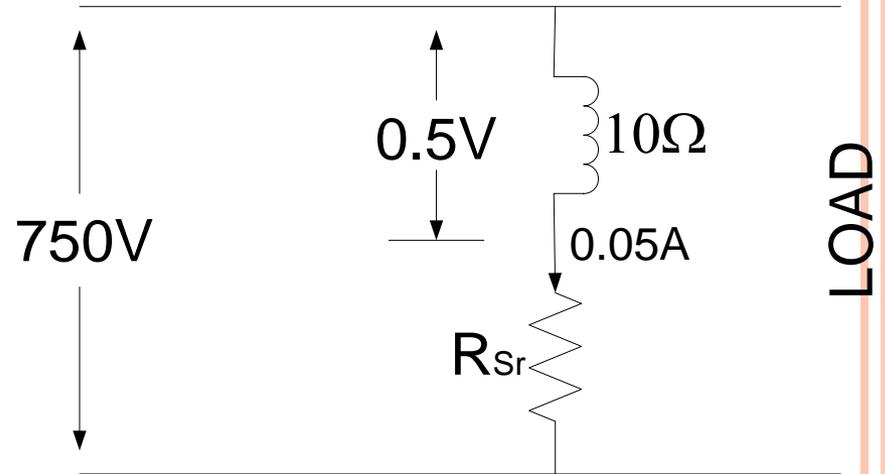
### *As ammeter:*

Current range can be extended by using a shunt resistor across the instrument.

Obviously,

$$10 \times 0.05 = R_S \times 99.95$$

$$\therefore R_S = 0.005\Omega$$



### *As voltmeter:*

The range can be extended by using a high resistance placed in series with the instrument,  $R_{Sr}$

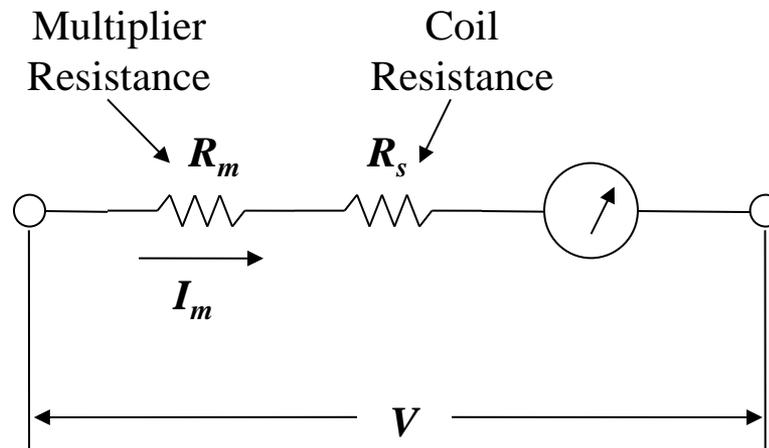
Obviously,  $R_{Sr}$  must drop a voltage of  $(750 - 0.5)V = 749.5V$  while carrying 0.05A

$$\therefore 0.05R_{Sr} = 749.5 \text{ or } R_{Sr} = 14.99k\Omega$$

## EXERCISE:

A PMMC instrument with FSD of  $100\mu\text{A}$  and a coil resistance of  $30\text{k}\Omega$  is to be converted into a voltmeter.

- Calculate the multiplier resistance required for the voltmeter to measure  $10\text{V}$  at full scale.
- Determine the applied voltage when the instrument indicates  $0.5$  FSD and  $0.1$  FSD.



## SOLUTION:

○ At  $V = 10V$  FSD

$$I_m = 100\mu A$$
$$R_s = \frac{V}{I_m} - R_m = \frac{10V}{100\mu A} - 30k\Omega = 70k\Omega$$

○ At  $V = 0.5$  FSD

$$I_m = 0.5 \times 100\mu A = 50\mu A$$

$$V = I_m (R_m + R_s) = 50\mu A (70k\Omega + 30k\Omega) = 5V$$

○ At  $V = 0.1$  FSD

$$I_m = 0.1 \times 100\mu A = 10\mu A$$

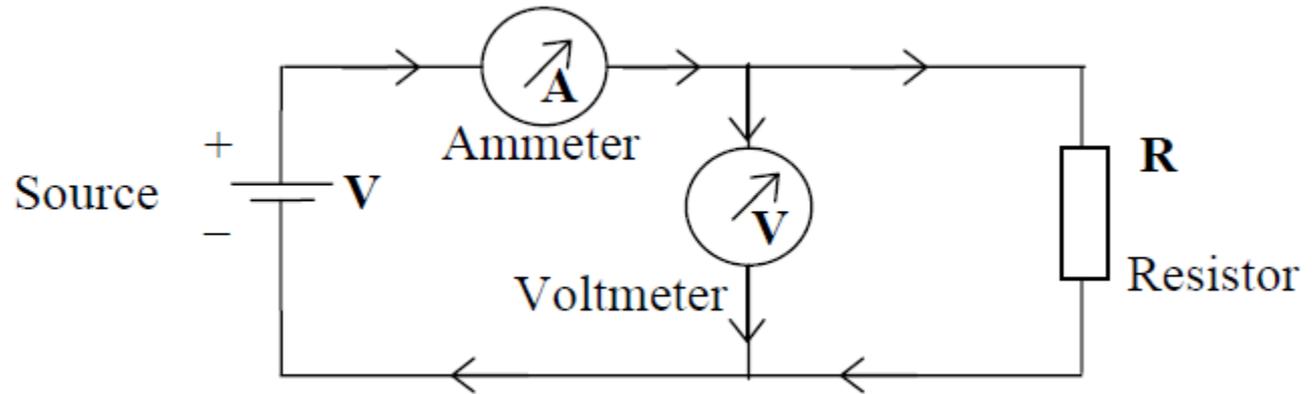
$$V = I_m (R_m + R_s) = 10\mu A (70k\Omega + 30k\Omega) = 1V$$

# **INDICATING INSTRUMENTS FOR RESISTANCE MEASUREMENTS**

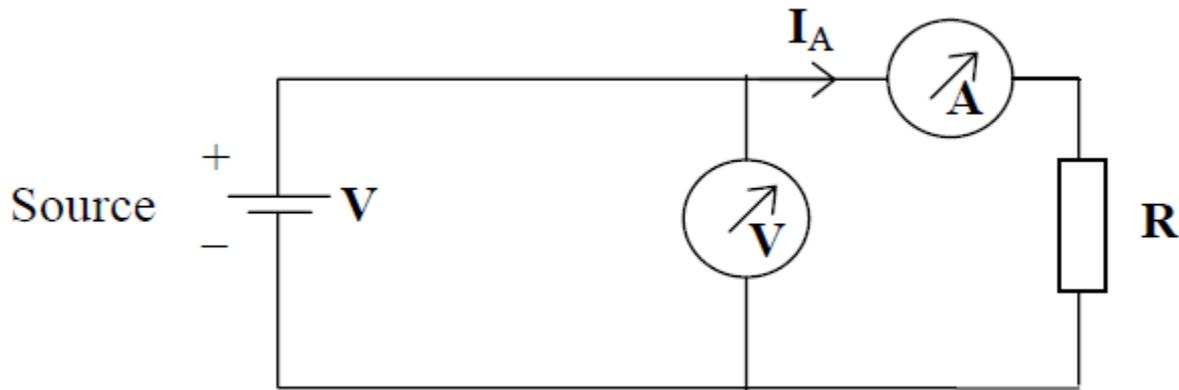
45

# Voltmeter Ammeter Method for Measuring Resistance

**Low R**



**High R**



# *Resistance*

the potential difference appearing across a device is proportional to the current flowing through it.

The connecting relationships is known as

Ohm's law and is written:

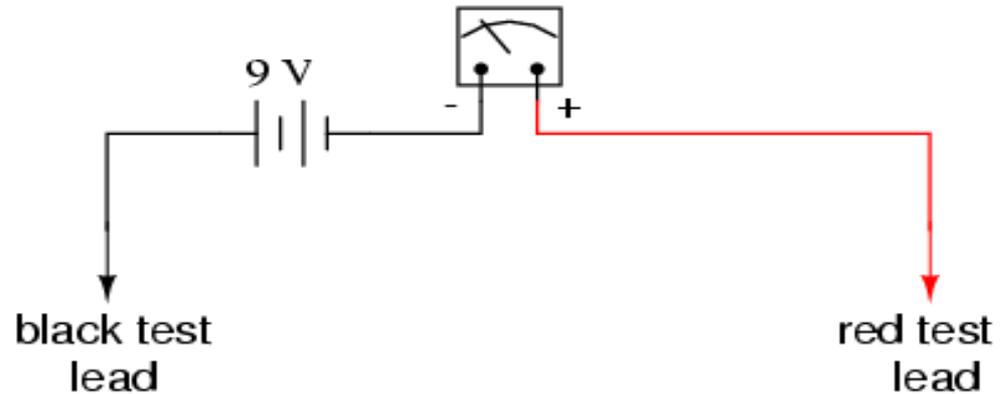
$$**V = IR**$$

where  $R$  is the resistance and is measured in ohms ( $\Omega$ ).

# Simple Voltmeter

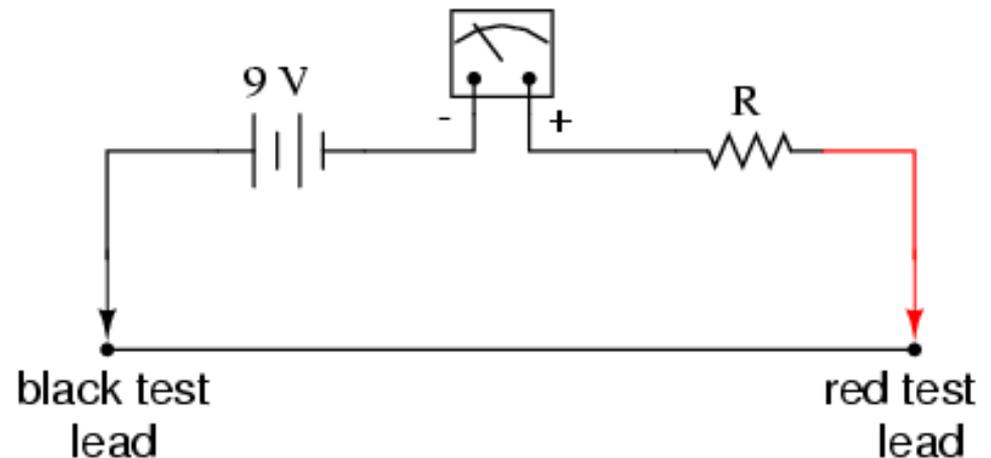
A simple ohmmeter

$500\ \Omega$  F.S. = 1 mA



If the test leads of this ohmmeter are directly **shorted** together (**measuring zero  $\Omega$** ), the meter movement will have a **maximum** amount of current through it, limited only by the battery **voltage** and the movement's internal **resistance**

$500\ \Omega$  F.S. = 1 mA



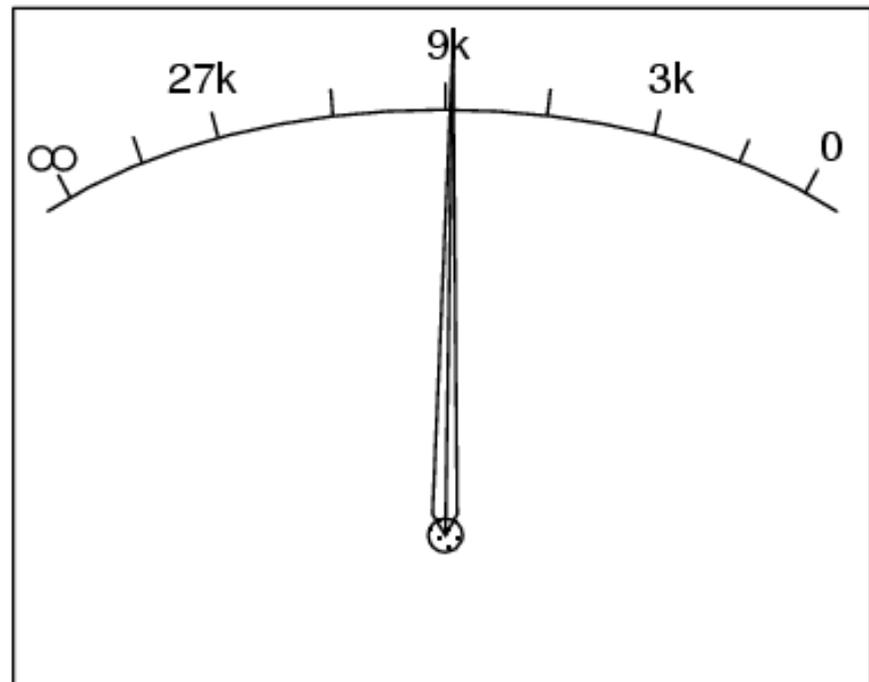
# Simple Voltmeter

To determine the proper **value for R**, we calculate the total circuit resistance needed to limit current to 1 mA (full-scale deflection on the movement) with 9 volts of potential from the battery, then subtract the movement's internal resistance from that figure:

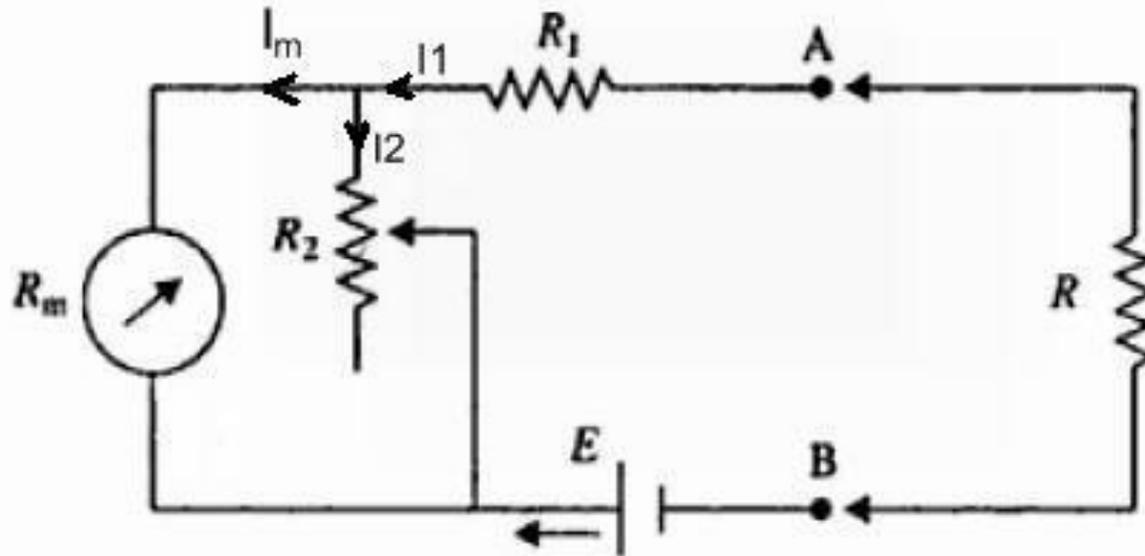
$$R_{\text{total}} = \frac{E}{I} = \frac{9 \text{ V}}{1 \text{ mA}}$$

$$R_{\text{total}} = 9 \text{ k}\Omega$$

$$R = R_{\text{total}} - 500 \Omega = 8.5 \text{ k}\Omega$$



# Series-Type Ohmmeter



Basic series type ohmmeter.

$R_1$  = current limiting resistor,

$R_2$  = **zero** adjusting resistor,

$E$  = emf of internal battery,

$R_m$  = internal resistance of d'Arsonval movement,

$R$  = the unknown resistor

# Series-Type Ohmmeter

- The current the the meter depends on the value of unknown resistor.
- Calibration problem should be taken into account.
- When  $R=0 \Omega$  (terminals A and B are shorted).
- $R=0 \Omega \rightarrow$  indicated the full-scale current  $I_{fsd}$
- $R= \infty$ , indicates zero current
- The disadvantage of this type is that battery voltage decreases with time and age. **Though not giving zero reading** when shorted.
- Change of the value of  $R_1$  : could change the calibration along the **scale**.
- The solution for battery aging is by **zero adjustment using  $R_2$** .

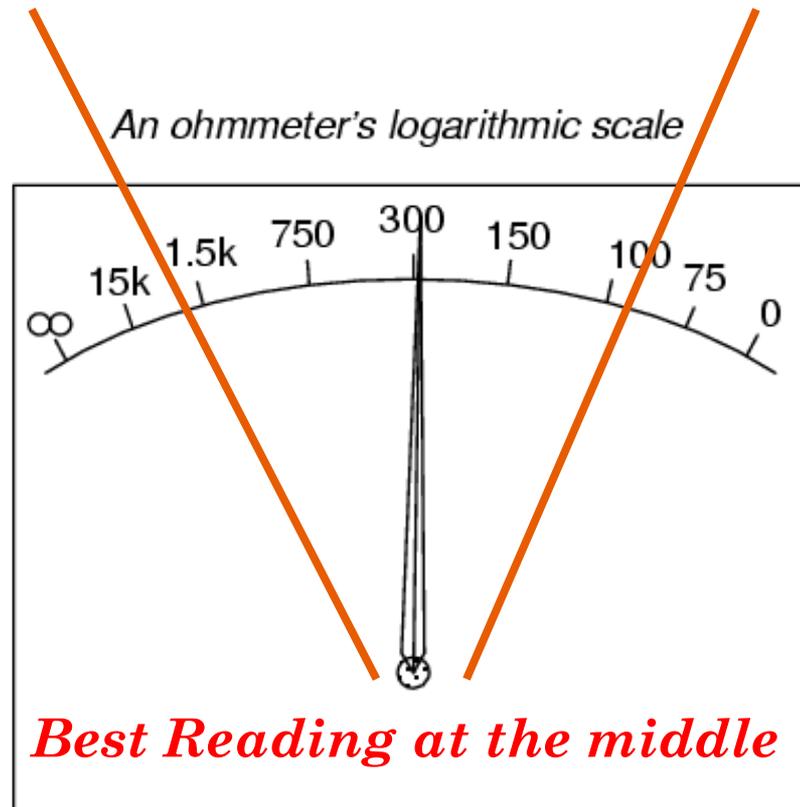
# *logarithmic scale*

The scale of an ohmmeter does not smoothly progress from **zero to infinity** as the needle sweeps from right to left.

The scale starts out "expanded" at the right-hand side, with the successive resistance values growing **closer and closer to each other toward the left** side of the scale

**Infinity cannot be approached in a linear (even) fashion**, because the scale would *never* get there!

With a logarithmic scale, **the amount of resistance spanned for any given distance on the scale** increases as the scale progresses toward infinity, making infinity an attainable goal.



- In the design of series-type ohmmeter, the design based on the value of **unknown** resistor (R) that cause **half-scale current**  $I_{h\text{sd}}$

$$R = R_h$$

- $R_h$  should be equal to the **total ohmmeter resistance**

$$R_h = R_1 + \frac{R_2 R_m}{R_2 + R_m}$$

- $R_h$  reduces meter current to  $0.5 I_{\text{fsd}}$

$$I_T = \frac{E}{2R_h} = 2I_h$$

- The total resistance presented to the battery

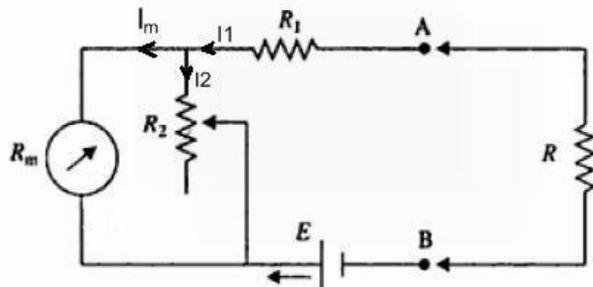
$$R_T = 2R_h$$

$$I_h = \frac{E}{2R_h}$$

$$V_{R_2} = V_m \implies$$

$$R_2 = \frac{I_{\text{fsd}} R_m}{I_T - I_{\text{fsd}}} = \frac{I_{\text{fsd}} R_m R_h}{E - I_{\text{fsd}} R_h}$$

$$R_1 = R_h - \frac{I_{\text{fsd}} R_m R_h}{E}$$



Basic series type ohmmeter.

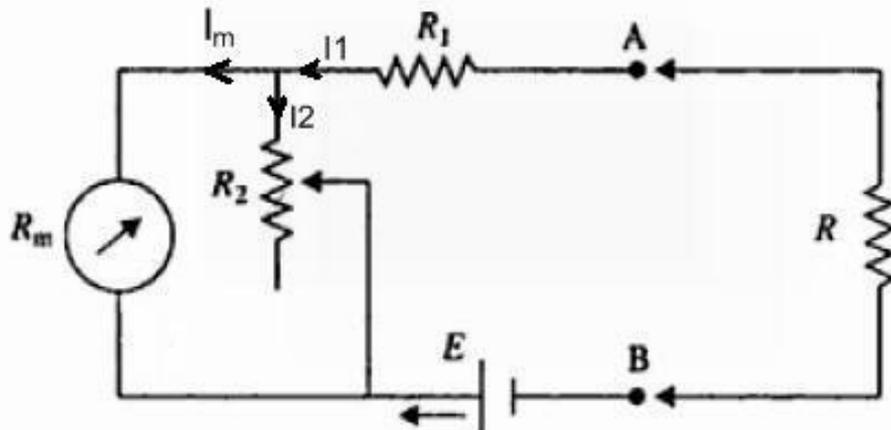
### Example 4-7

The ohmmeter of Fig. 4-21 uses a  $50\text{-}\Omega$  basic movement requiring a full-scale current of  $1\text{ mA}$ . The internal battery voltage is  $3\text{ V}$ . The desired scale marking for half-scale deflection is  $2,000\ \Omega$ . Calculate (a) the values of  $R_1$  and  $R_2$ ; (b) the maximum value of  $R_2$  to compensate for a  $10\%$  drop in battery voltage; (c) the scale error at the half-scale mark ( $2,000\ \Omega$ ) when  $R_2$  is set as in (b).

### Solution

(a) The total battery current at full-scale deflection is **( $R_x = 0\ \Omega$ )**

$$I_t = \frac{E}{R_h} = \frac{3\text{ V}}{2,000\ \Omega} = 1.5\text{ mA} \quad (4-16)$$



Basic series type ohmmeter.

The current through the zero-adjust resistor  $R_2$  then is

$$I_2 = I_f - I_{fsd} = 1.5 \text{ mA} - 1 \text{ mA} = 0.5 \text{ mA}$$

The value of the zero-adjust resistor  $R_2$  is

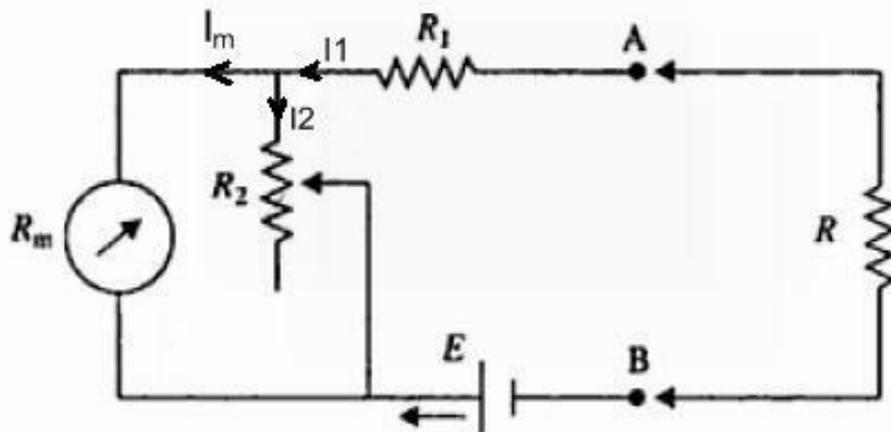
$$R_2 = \frac{I_{fsd} R_m}{I_2} = \frac{1 \text{ mA} \times 50 \Omega}{0.5 \text{ mA}} = 100 \Omega$$

The parallel resistance of the movement and the shunt ( $R_p$ ) is

$$R_p = \frac{R_2 R_m}{R_2 + R_m} = \frac{50 \times 100}{150} = 33.3 \Omega$$

The value of the current-limiting resistor  $R_1$  is

$$R_1 = R_h - R_p = 2,000 - 33.3 = 1,966.7 \Omega$$



(b) At a 10% drop in battery voltage,

$$E = 3 \text{ V} - 0.3 \text{ V} = \underline{2.7 \text{ V}}$$

The total battery current  $I_t$  then becomes

$$I_t = \frac{E}{R_h} = \frac{2.7 \text{ V}}{2,000 \Omega} = 1.35 \text{ mA}$$

The shunt current  $I_2$  is

$$I_2 = I_t - I_{fsd} = 1.35 \text{ mA} - 1 \text{ mA} = 0.35 \text{ mA}$$

and the zero-adjust resistor  $R_2$  equals

$$R_2 = \frac{I_{fsd} R_m}{I_2} = \frac{1 \text{ mA} \times 50 \Omega}{0.35 \text{ mA}} = \underline{143 \Omega}$$

(c) The parallel resistance of the meter movement and the new value of  $R_2$  becomes

$$R_p = \frac{R_2 R_m}{R_2 + R_m} = \frac{50 \times 143}{193} = 37 \Omega$$

Since the half-scale resistance  $R_h$  is equal to the total internal circuit resistance,  $R_h$  will increase to

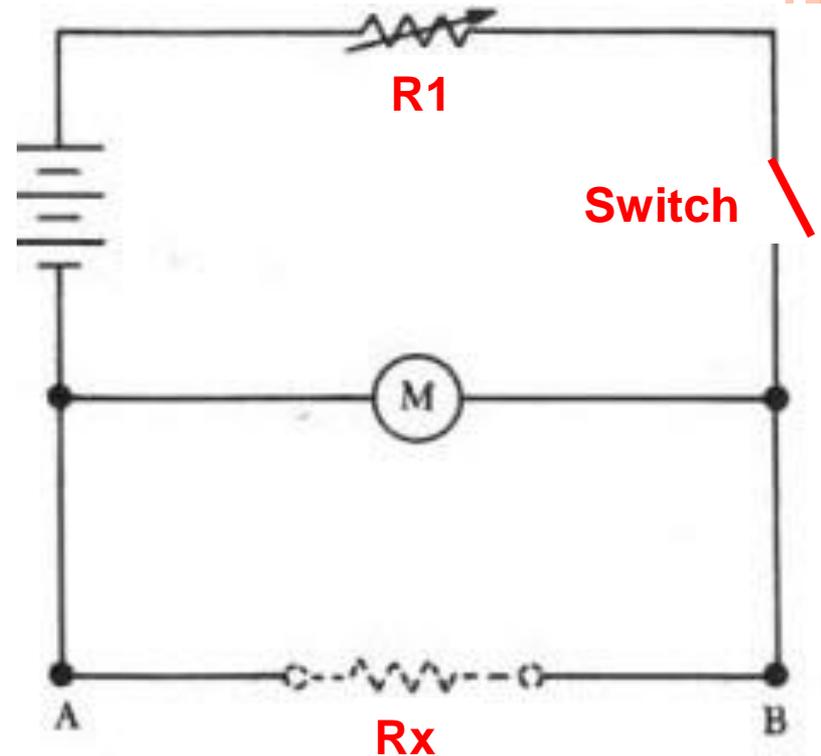
$$R_h = R_1 + R_p = 1,966.7 \Omega + 37 \Omega = 2,003.7 \Omega$$

Therefore the true value of the half-scale mark on the meter is 2,003.7  $\Omega$  whereas the actual scale mark is 2,000  $\Omega$ . The percentage error is then

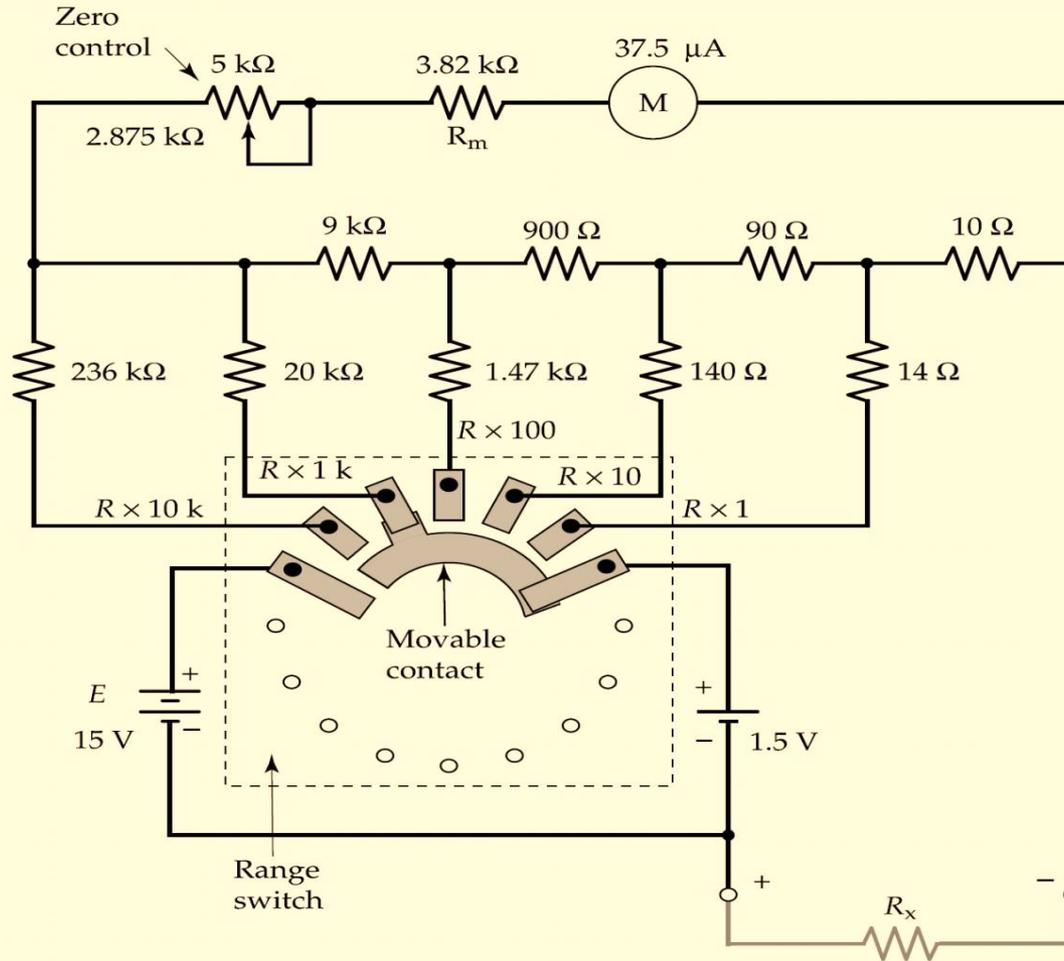
$$\% \text{ error} = \frac{2,000 - 2,003.7}{2,007.3} \times 100\% = -0.185\%$$

# Shunt-Type Ohmmeter

- Not commonly used.
- The shunt type mainly used for measuring **low vale** resistors.
- *ON/Off* Switch is placed to connect and disconnect the battery.
- The full scale reading depends on  $R_1$  and  $R_m$ .
- Designed same as the same the series type based on half-scale reading  **$R_h$** .

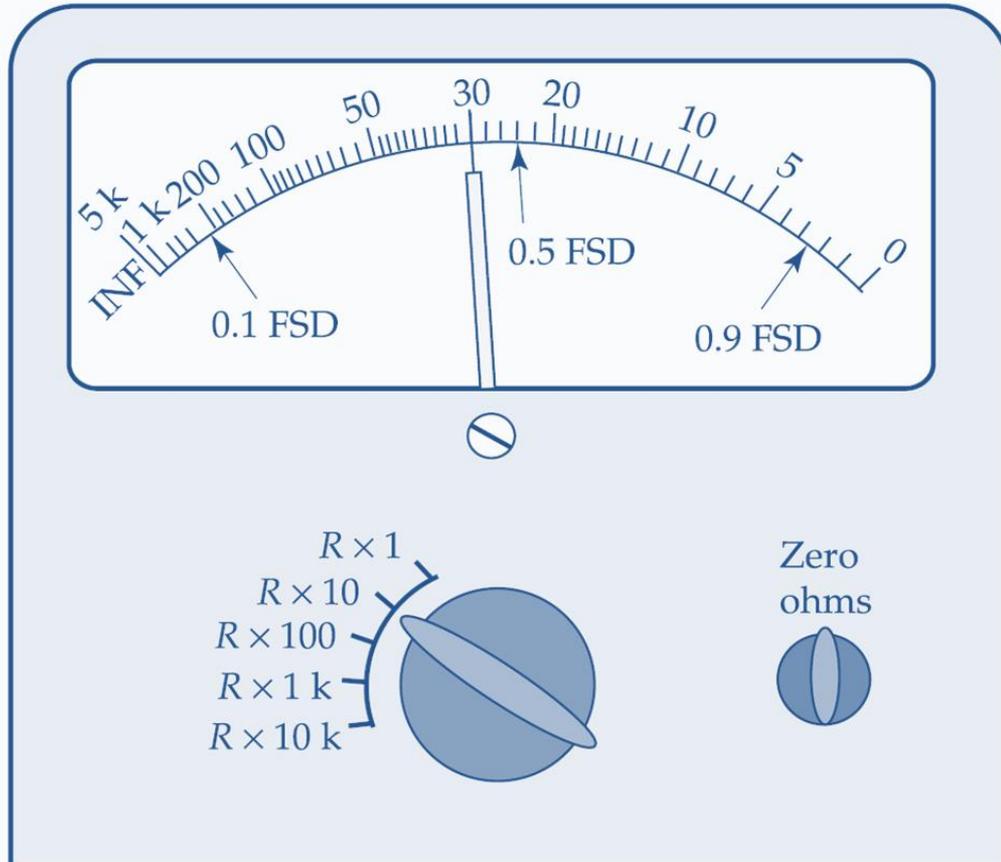


# Multirange Ohmmeter Circuit



**Figure 4-13** Circuit for a typical multirange shunt ohmmeter as used on a multifunction analog instrument. The 15 V battery is used only on the  $R \times 10$  k $\Omega$  range, and the 1.5 V battery is the supply for all other ranges.

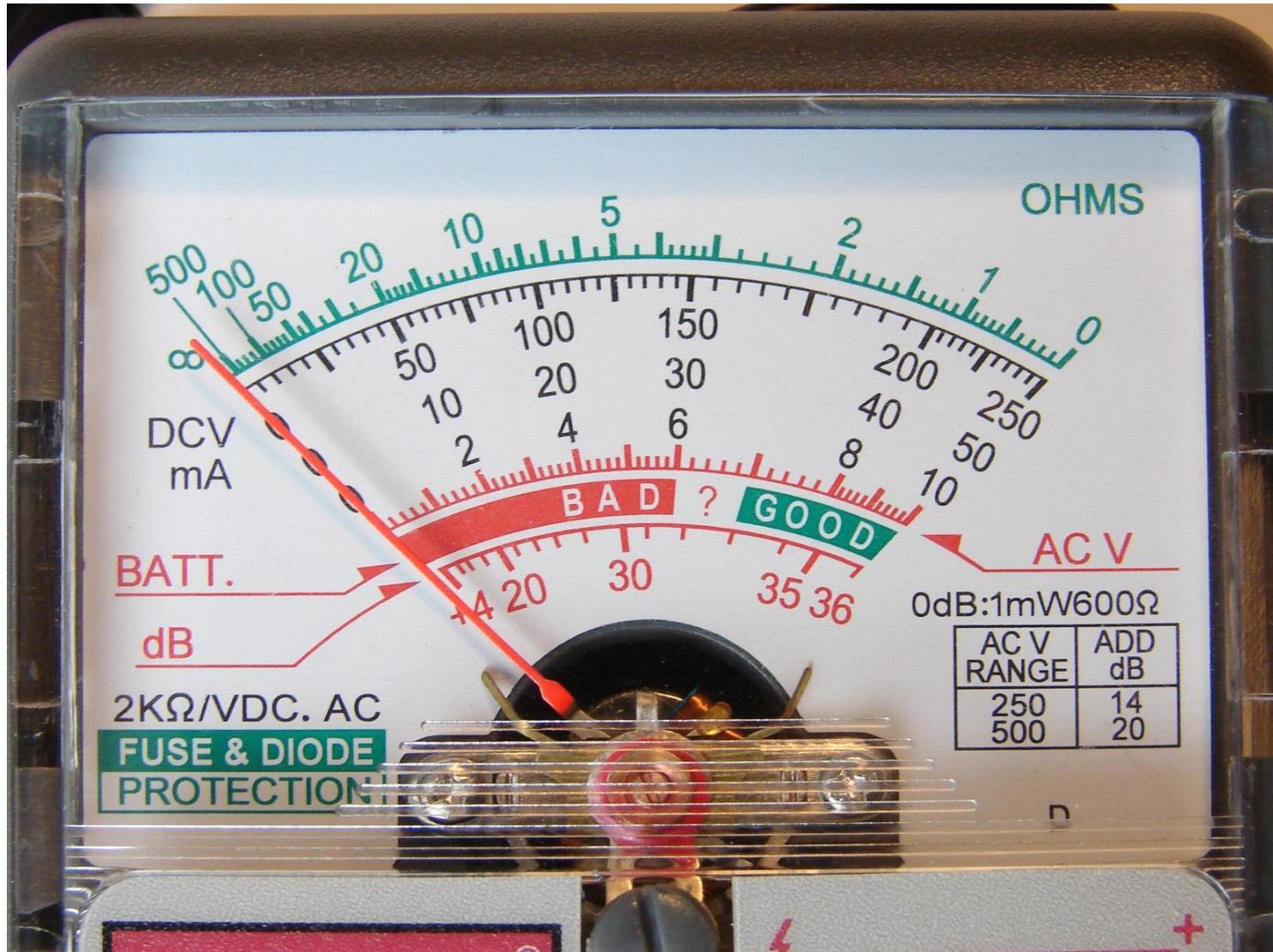
# Scale & Range Switch for a Typical Multirange Ohmmeter

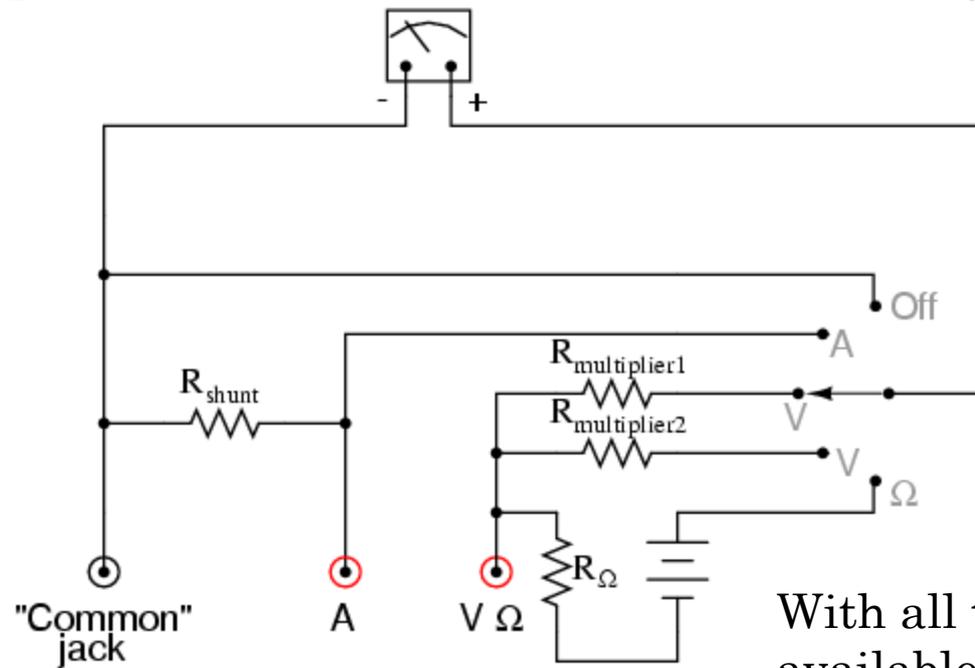
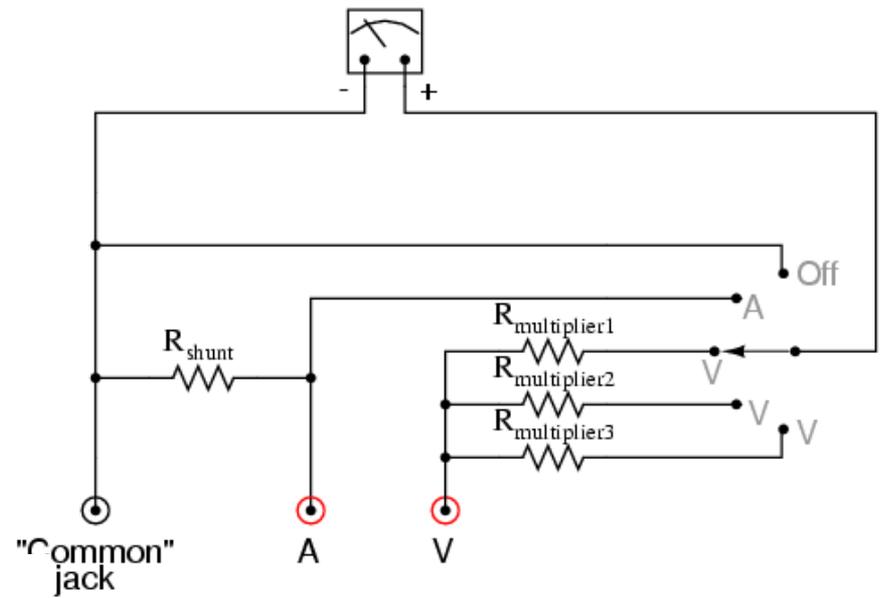


**Figure 4-14** Scale and range switch for a typical multirange shunt ohmmeter as used on an analog VOM.

# Multimeters

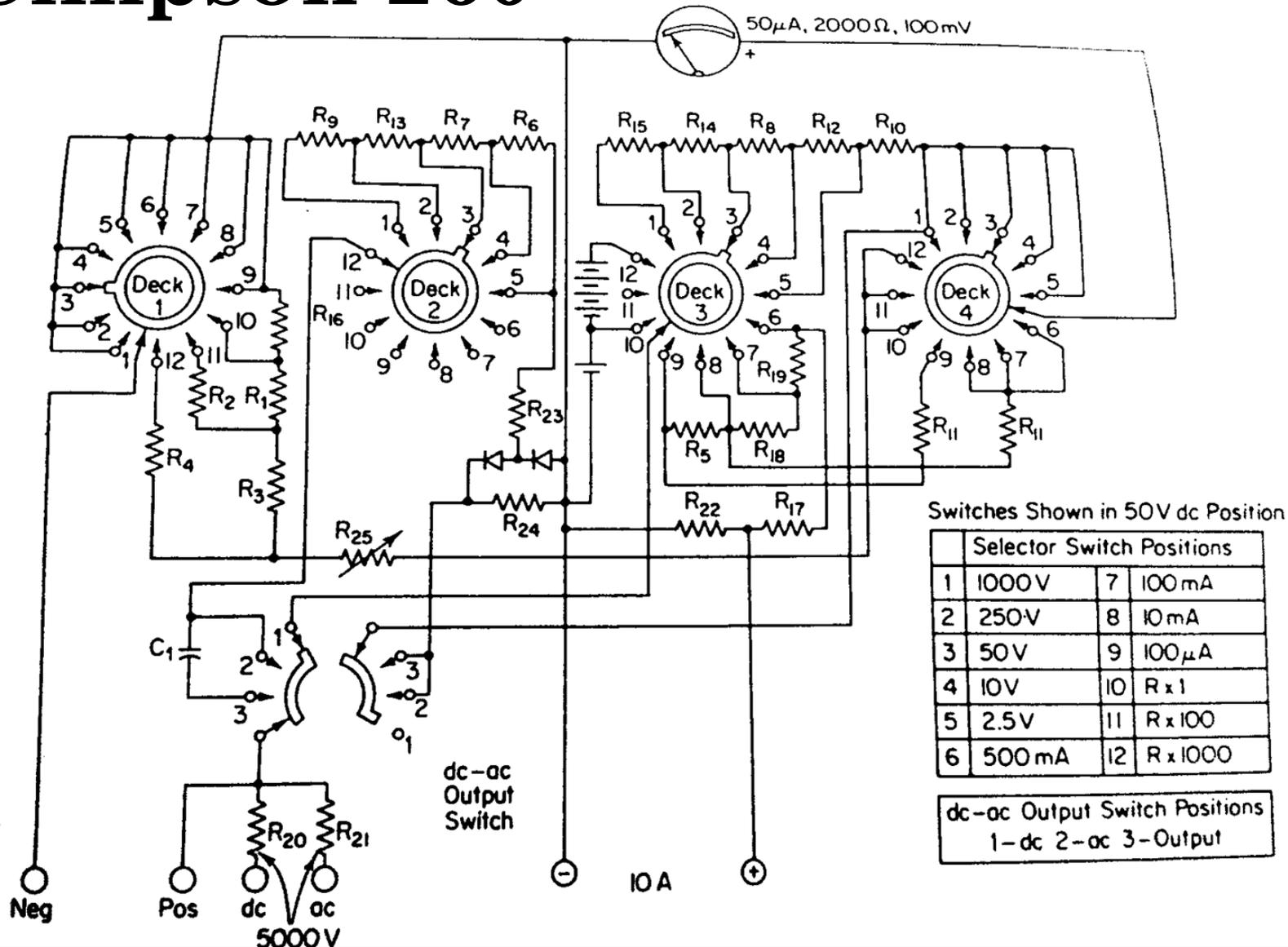
volt - ohm - Ampere





With all three fundamental functions available, this multimeter may also be known as a *volt-ohm-milliammeter*.

# Multimeter Example : Simpson 260



# Multimeter Example : Simpson 260

Fig 4-10 Multimeter or VOM

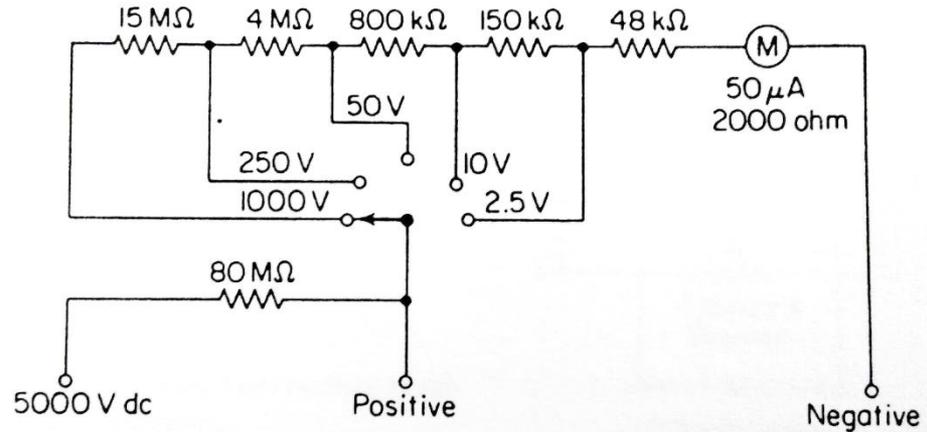


Figure 4-25 Dc voltmeter section of the Simpson Model 260 multimeter (courtesy Simpson Electric Company).

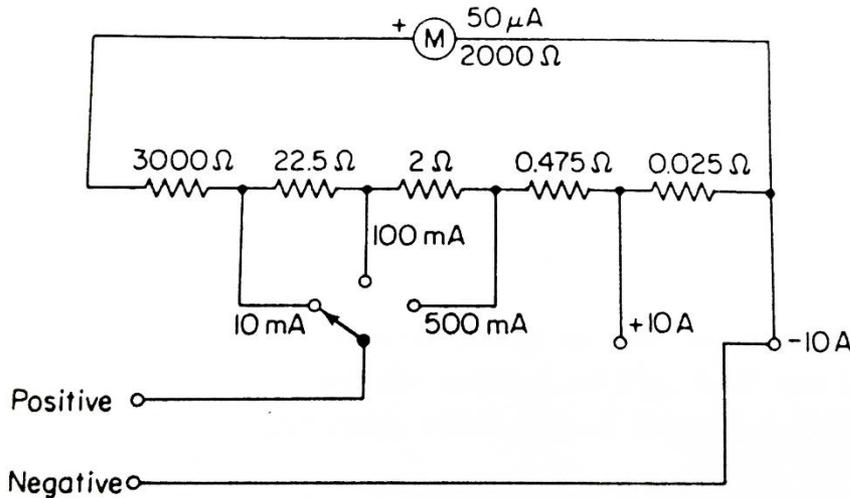
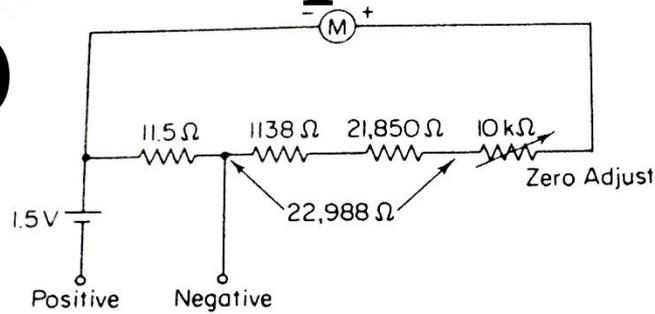
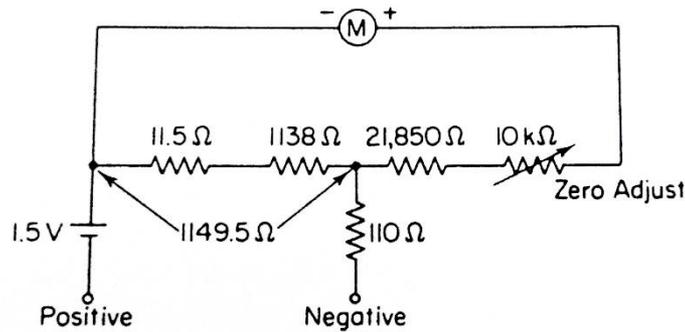


Figure 4-26 Dc ammeter section of the Simpson Model 260 multimeter (courtesy Simpson Electric Company).

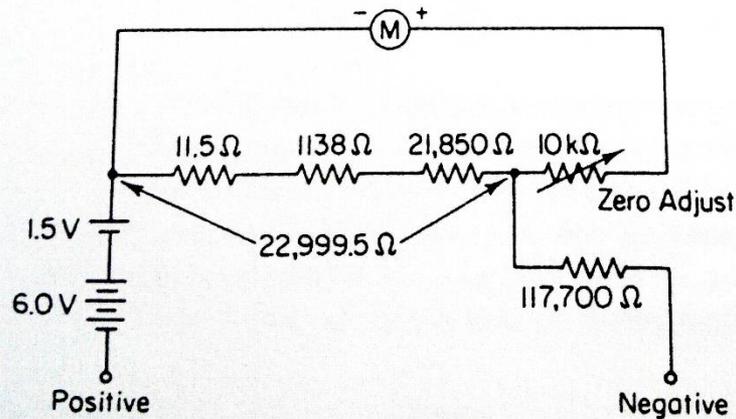
# Multimeter Example : Simpson 260



(a) Ohmmeter Circuit Rx1 Range



(b) Ohmmeter Circuit Rx100 Range



(c) Ohmmeter Circuit Rx10,000 Range

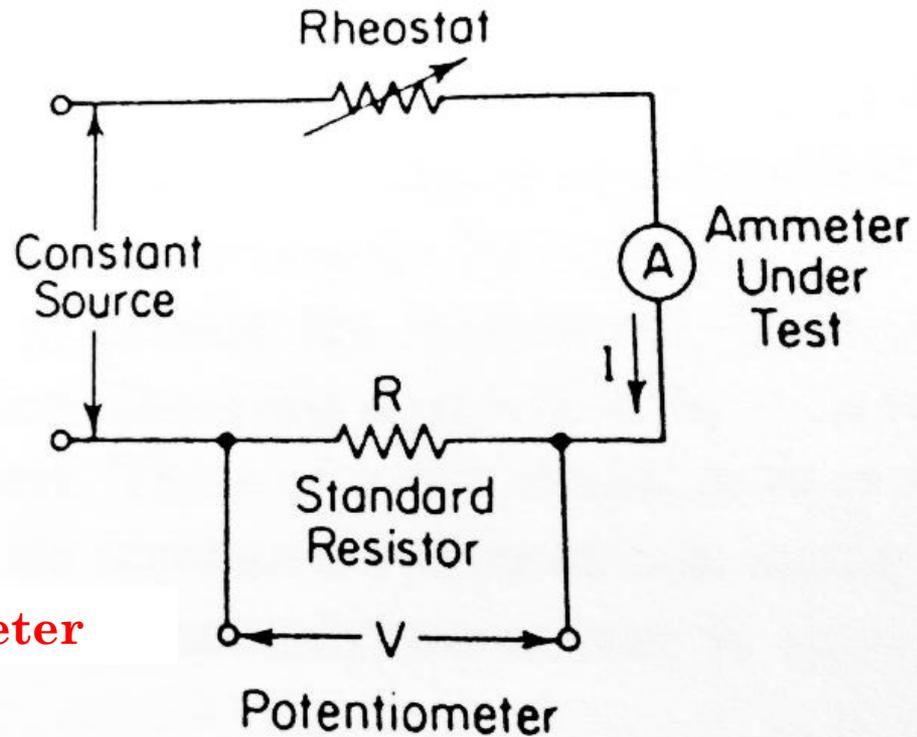
# Calibration of DC Instruments (**Ammeter**)

## **DC Ammeter** Calibration

Based on ohm's Law

**Compare the current calculated with the meter reading.**

**DC Ammeter**



# Calibration of DC Instruments (**Voltmeter**)

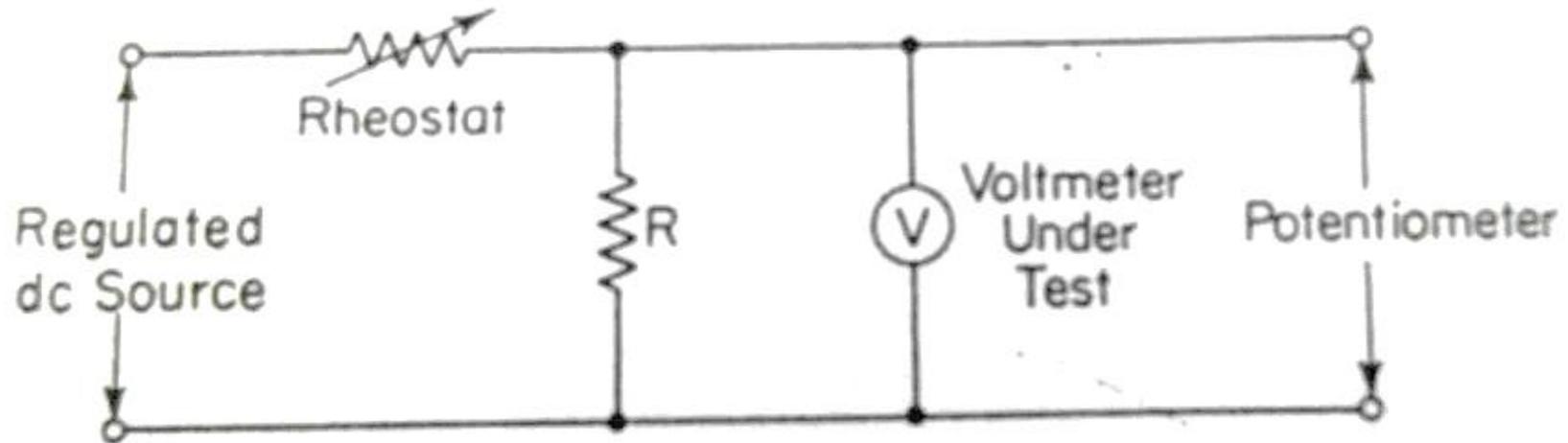


Figure 4-29 Potentiometer method of calibrating a dc voltmeter.

**The potentiometer voltage should be = Meter voltage**

# Calibration of DC Instruments (**Ohm Meter**)

Ohm meter calibrated using  
standard resistor.

*First Exam on*  
**30/03/2015**

